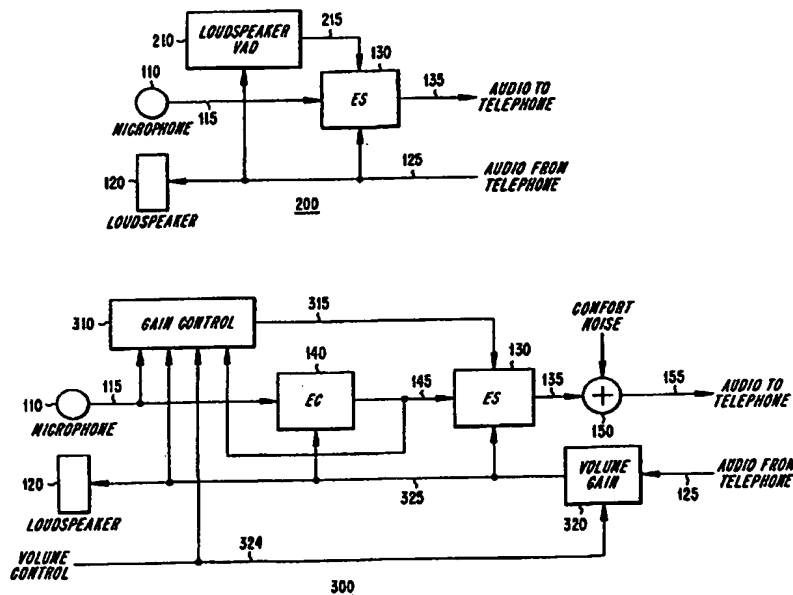


PCTWORLD INTELLECTUAL PROPERTY ORGANIZATION
International Bureau

INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification ⁶ : H04M 9/08	A1	(11) International Publication Number: WO 99/35812 (43) International Publication Date: 15 July 1999 (15.07.99)
(21) International Application Number: PCT/US98/26231 (22) International Filing Date: 22 December 1998 (22.12.98) (30) Priority Data: 09/005,144 9 January 1998 (09.01.98) US (71) Applicant: ERICSSON INC. [US/US]; 7001 Development Drive, P.O. Box 13969, Research Triangle Park, NC 27709 (US). (72) Inventor: ROMESBURG, Eric, Douglas; 18 Rocky Knolls Road, Chapel Hill, NC 27516 (US). (74) Agents: GRUDZIECKI, Ronald, L. et al.; Burns, Doane, Swecker & Mathis, L.L.P., P.O. Box 1404, Alexandria, VA 22313-1404 (US).		(81) Designated States: AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, UA, UG, UZ, VN, YU, ZW, ARIPO patent (GH, GM, KE, LS, MW, SD, SZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG). Published <i>With international search report.</i>
(54) Title: METHODS AND APPARATUS FOR CONTROLLING ECHO SUPPRESSION IN COMMUNICATIONS SYSTEMS (57) Abstract <p>Methods and apparatus for controlling echo suppressors provide superior echo attenuation while optimizing perceived signal quality under a variety of conditions. The disclosed techniques provide desired levels of echo attenuation while minimizing the likelihood that an echo suppressor will introduce more attenuation, and thus more signal distortion, than is necessary. In a first exemplary embodiment, an acoustic echo suppressor is deactivated when a voice activity detector indicates that a far-end communications signal does not include a speech component. In a second exemplary embodiment, an attenuation factor of an active acoustic echo canceler is adjusted based upon whether a voice activity detector indicates that a near-end communications signal includes a near-end speech component. In a third exemplary embodiment, a level of attenuation provided by an acoustic echo canceler is adjusted based upon a measured ratio of voice to noise in an audio input to the acoustic echo canceler.</p>		



FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	Albania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
BB	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav Republic of Macedonia	TM	Turkmenistan
BF	Burkina Faso	GR	Greece			TR	Turkey
BG	Bulgaria	HU	Hungary	ML	Mali	TT	Trinidad and Tobago
BJ	Benin	IE	Ireland	MN	Mongolia	UA	Ukraine
BR	Brazil	IL	Israel	MR	Mauritania	UG	Uganda
BY	Belarus	IS	Iceland	MW	Malawi	US	United States of America
CA	Canada	IT	Italy	MX	Mexico	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NE	Niger	VN	Viet Nam
CG	Congo	KE	Kenya	NL	Netherlands	YU	Yugoslavia
CH	Switzerland	KG	Kyrgyzstan	NO	Norway	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's Republic of Korea	NZ	New Zealand		
CM	Cameroon			PL	Poland		
CN	China	KR	Republic of Korea	PT	Portugal		
CU	Cuba	KZ	Kazakhstan	RO	Romania		
CZ	Czech Republic	LC	Saint Lucia	RU	Russian Federation		
DE	Germany	LI	Liechtenstein	SD	Sudan		
DK	Denmark	LK	Sri Lanka	SE	Sweden		
EE	Estonia	LR	Liberia	SG	Singapore		

METHODS AND APPARATUS FOR CONTROLLING ECHO SUPPRESSION IN COMMUNICATIONS SYSTEMS

BACKGROUND OF THE INVENTION

5 The present invention relates to communications systems, and more particularly,
to echo suppression in bi-directional communications links.

In many communications systems, for example landline and wireless telephone
systems, voice signals are often transmitted between two system users via a bi-
directional communications link. In such systems, speech of a near-end user is
typically detected by a near-end microphone at one end of the communications link and
10 then transmitted over the link to a far-end loudspeaker for reproduction and
presentation to a far-end user. Conversely, speech of the far-end user is detected by a
far-end microphone and then transmitted via the communications link to a near-end
loudspeaker for reproduction and presentation to the near-end user. At either end of
the communications link, loudspeaker output detected by a proximate microphone may
15 be inadvertently transmitted back over the communications link, resulting in what may
be unacceptably disruptive feedback, or echo, from a user perspective.

Therefore, in order to avoid transmission of such undesirable echo signals, the
microphone acoustic input should be isolated from loudspeaker output as much as
possible. With a conventional telephone handset, in which the handset microphone is
20 situated close to the user's mouth while the handset speaker essentially covers the
user's ear, the requisite isolation is easily achieved. However, as the physical size of
portable telephones has decreased, and as hands-free speaker-phones have become more
popular, manufacturers have moved toward designs in which the acoustic path from the
loudspeaker to the microphone is not blocked by the user's head or body. As a result,
25 the need for more sophisticated echo suppression techniques has become paramount in
modern systems.

The need is particularly pronounced in the case of hands-free automobile
telephones, where the closed vehicular environment can cause multiple reflections of a
loudspeaker signal to be coupled back to a high-gain hands-free microphone.
30 Movement of the user in the vehicle and changes in the relative directions and strengths

of the echo signals, for example as windows are opened and closed or as the user moves his head while driving, further complicate the task of echo suppression in the automobile environment. Additionally, more recently developed digital telephones process speech signals through voice encoders which introduce significant signal delays and create non-linear signal distortions. Such prolonged delays tend to magnify the problem of signal echo from a user perspective, and the additional non-linear distortions make echo suppression by the network equipment more difficult.

In response to the above described challenges, telephone manufacturers have developed a wide variety of echo suppression mechanisms. An exemplary echo suppression system 100 is depicted in Figure 1A. As shown, the exemplary system 100 includes a microphone 110, a loudspeaker 120 and an echo suppressor 130. An audio output 115 of the microphone 110 is coupled to an audio input of the echo suppressor 130, and an audio output 135 of the echo suppressor 130 serves as a near-end audio input to a telephone (not shown). Additionally, a far-end audio output 125 from the telephone is coupled to an audio input of the loudspeaker 120 and to a reference input of the echo suppressor 130.

In operation, the echo suppressor 130 processes the microphone signal 115 to provide the audio output signal 135 to a far-end telephone user. More specifically, the echo suppressor 130 attenuates the microphone signal 115, in dependence upon the far-end audio signal 125, so that acoustic echo from the loudspeaker 120 to the microphone 110 is not passed back to the far-end telephone user.

Typically, the echo suppressor 130 is either a non-linear, clipping type suppressor or a linear, scaling type suppressor. Clipping type suppressors generally attenuate the microphone output signal 115 by removing a portion of the signal falling within a particular range of values (i.e., within a particular clipping window). Scaling type suppressors, on the other hand, attenuate the microphone output signal 115 by multiplying the signal with an appropriate scale factor. Recently developed hybrid suppressors incorporate both clipping and scaling aspects, for example by scaling a portion of the microphone signal falling within a particular attenuation window. In any case, the level of attenuation (i.e., the clipping window and/or the scale factor) is

generally adjusted, either directly or indirectly, in accordance with the amplitude of the far-end audio signal 125 so that the microphone output 115 is attenuated only to the extent the far-end user is speaking.

5 A conventional clipping type suppressor, known in the art as a center clipper, is described for example in U.S. Patent No. 5,475,731, entitled "Echo-Canceling System and Method Using Echo Estimate to Modify Error Signal" and issued December 12, 1995 to Rasmusson et al. An alternative clipping type suppressor, known as an AC-Center clipper, is described in copending U.S. Patent Application No. 08/775,797, entitled "An AC-Center Clipper for Noise and Echo Suppression in a Communications
10 System" and filed December 31, 1996. An exemplary scaling type suppressor is described in U.S. Patent No. 5,283,784, entitled "Echo Canceller Processing Techniques and Processing" and issued February 1, 1994 to Genter. An advanced hybrid suppressor, referred to herein as an AC-center attenuator, is described in copending U.S. Patent Application 09/005,149, entitled "Methods and Apparatus for
15 Improved Echo Suppression in Communications Systems" and filed on even date herewith. Each of the above identified patents, as well as each of the above identified patent applications, is incorporated herein in its entirety by reference.

The echo suppressor 130 of Figure 1A can also be combined with a linear echo canceler and/or a comfort noise generator to provide a more sophisticated echo
20 suppression system. Figure 1B depicts an exemplary system 101 including the microphone 110, the loudspeaker 120 and the echo suppressor 130 of Figure 1A, and an acoustic echo canceler 140 and a comfort noise summing device 150. As shown, the microphone output 115 is coupled to an audio input of the acoustic echo canceler 140, and an audio output 145 of the acoustic echo canceler 140 is coupled to the audio
25 input of the echo suppressor 130. The suppressor audio output 135 and a comfort noise signal are coupled to first and second inputs of the summing device 150, respectively, and an audio output 155 of the summing device 150 serves as the near-end audio input to the telephone (not shown). The far-end audio output 125 from the
30 of the acoustic echo canceler 140 and the echo suppressor 130.

In operation, the acoustic echo canceler 140 dynamically models the acoustic path from the loudspeaker 120 to the microphone 110 and attempts to cancel, from the microphone output signal 115, any loudspeaker sound that is picked up by the microphone 110. Algorithms commonly used for modeling the acoustic echo path include the well known Least Mean Squares (LMS) algorithm and variants such as Normalized Least Mean Squares (NLMS). An exemplary Least Mean Squares based canceler is described in the above cited U.S. Patent No. 5,475,731 to Rasmusson et al. Additionally, an advanced Normalized Least Mean Squares based canceler is described in copending U.S. Patent Application No. 08/852,729, entitled "An Improved Echo Canceler for use in Communications Systems" and filed May 7, 1997, which is incorporated herein in its entirety by reference.

In the system 101 of Figure 1B, the echo suppressor 130 supplements the echo canceler 140 and provides additional echo attenuation as necessary. For example, the echo suppressor 130 can be dynamically adjusted based on a measurement of the instantaneous level of echo cancelation achieved by the acoustic echo canceler 140 to attenuate the residual echo to a predetermined goal level. Advanced methods for dynamically measuring the echo cancelation provided by an echo canceler are described, for example, in the above cited U.S. Patent Application No. 09/005,149 (entitled "Methods and Apparatus for Improved Echo Suppression in Communications Systems" and filed on even date herewith).

As in the system 100 of Figure 1A, the echo suppressor 130 can be a clipping suppressor, a scaling suppressor or a hybrid suppressor. Additionally, the echo suppressor 130 can, when following the echo canceler 140, be a simple single-throw switch which selectively cuts out, or mutes, the audio output signal 135 at appropriate times (e.g., during periods in which a voice activity detector indicates that the microphone signal 115 contains no speech). In any case, the echo suppressor 130 attenuates the entire audio signal and, in addition to suppressing echo, distorts any background noise and/or near-end speech which may be present. In fact, the background noise can be suppressed to the point that the far-end user may erroneously believe that the call has been disconnected when the echo suppressor 130 is active.

Therefore, to improve the quality of communication for the far-end user, today's systems often add comfort noise to the telephone audio signal 135 when the echo suppressor 130 is active. In the system of Figure 1B, comfort noise is added to the suppressor output signal 135 via the summing device 150. Alternately, the summing device 150 can be replaced with a switch so that either the suppressor output signal 135 or the comfort noise signal is selectively passed to the far-end user. Advanced methods for generating and utilizing comfort noise are described, for example, in U.S. Patent Application No. 09/005,145, entitled "Methods and Apparatus for Providing Comfort Noise in Communications Systems" and filed on even date herewith, which is incorporated herein in its entirety by reference.

While the above described systems provide sophisticated echo suppression, known methods for controlling such systems are inadequate in several respects. For example, known methods for controlling the echo suppressor 130 often result in the far-end user perceiving unacceptably distorted near-end speech. In other words, known methods of control sometimes activate the echo suppressor 130 when it should not be activated (e.g., during near-end single talk) and/or cause the echo suppressor 130 to attenuate the near-end signal more than is necessary or desirable (e.g., during periods of double talk). Additionally, certain known methods of controlling a residual echo suppressor (e.g., closing and opening a single-throw switch which follows a front-end echo canceler in dependence upon the presence or absence of near-end speech) can sometimes provide inadequate echo suppression. Consequently, there is a need for improved methods and apparatus for controlling echo suppressors.

SUMMARY OF THE INVENTION

The present invention fulfills the above-described and other needs by providing methods and apparatus for controlling an echo suppressor to provide a desired level of echo attenuation while minimizing the likelihood that the echo suppressor will introduce more attenuation, and thus more distortion, than is necessary at any given time. Advantageously, the methods and apparatus of the present invention permit

construction of communications systems which provide excellent echo suppression while optimizing perceived signal quality under a variety of conditions.

5 In a first exemplary embodiment, an echo suppression device according to the invention includes an echo suppressor configured to attenuate a first communications signal (e.g., a near-end audio signal). More specifically, the echo suppressor attenuates the first communications signal in order to suppress an echo component of the first communications signal, wherein the echo component results from a reflection of a second communications signal (e.g., a far-end audio signal). The exemplary echo suppression device also includes a voice activity detector configured to provide an indication of whether the second communications signal includes a speech component, and the echo suppressor is deactivated when the voice activity detector indicates that the second communications signal does not include an echo-causing speech component. As a result, the echo suppressor is active (and introduces distortion on the first communications signal) only when necessary.

15 In a second exemplary embodiment, an echo suppression device according to the invention includes an echo suppressor configured to multiply a portion of a first communications signal by an attenuation factor in order to suppress an echo component thereof, the echo component resulting from a reflection of a second communications signal. The exemplary echo suppression device also includes a gain control processor configured to detect a prevailing condition of the communications signals and to adjust the suppressor attenuation factor accordingly. For example, the gain control processor can set the suppressor attenuation factor to one of two values in dependence upon whether the first communications signal includes a near-end speech component. As a result, the exemplary echo suppression device can provide an appropriate level of echo attenuation while introducing minimal signal distortion.

20 In a third exemplary embodiment, an echo suppression device according to the invention includes an echo suppressor configured to attenuate a first communications signal in order to suppress an echo component thereof, the echo component resulting from a reflection of a second communications signal. The echo suppression device also includes a gain control processor configured to detect a ratio of voice to noise in the

30

first communications signal and to adjust a level of attenuation provided by the echo suppressor based upon the detected ratio. For example, the gain control processor can deactivate the echo suppressor when the ratio is at or below a predefined threshold. Alternatively, the gain control processor can adjust an attenuation factor of the echo suppressor to ensure that a ratio of echo to noise in an output of the echo suppression device is at or below a predefined threshold. As a result, the exemplary echo suppression device can provide an appropriate level of overall echo attenuation while introducing a minimum of signal distortion.

The above-described and other features of the present invention are explained in detail hereinafter with reference to the illustrative examples shown in the accompanying drawings. Those skilled in the art will appreciate that the described embodiments are provided for purposes of illustration and understanding and that numerous equivalent embodiments are contemplated herein.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1A depicts an exemplary echo suppression system in which the teachings of the present invention can be implemented.

Figure 1B depicts an alternative exemplary echo suppression system in which the teachings of the present invention can be implemented.

Figure 2 depicts an exemplary echo suppression system according to the invention, the exemplary system including a voice activity detector configured to provide a control input to an acoustic echo suppressor.

Figure 3 depicts an alternative echo suppression system according to the invention, the alternative system including a gain control processor configured to adjust an attenuation factor of an acoustic echo suppressor.

Figure 4 depicts steps in an exemplary method of controlling an echo suppressor according to the invention.

DETAILED DESCRIPTION OF THE INVENTION

Figure 2 depicts an exemplary echo suppression system 200 constructed in accordance with a first aspect of the invention. As shown, the exemplary system 200 includes the microphone 110, the loudspeaker 120 and the echo suppressor 130 of Figures 1A and 1B, and a loudspeaker voice activity detector 210. The audio output 115 of the microphone 110 is coupled to the audio input of the echo suppressor 130, and the audio output 135 of the echo suppressor 130 serves as the near-end audio input to the telephone (not shown). The far-end audio output 125 from the telephone is coupled to the audio input of the loudspeaker 120 and to reference inputs of the echo suppressor 130 and the voice activity detector 210. A binary output 215 of the voice activity detector 210 is coupled to a control input of the echo suppressor 130.

Generally, operation of the system 200 of Figure 2 is similar to that of the system 100 of Figure 1A. In other words, the echo suppressor 130 attenuates the microphone signal 115, in dependence upon the far-end audio signal 125, so that acoustic echo from the loudspeaker 120 to the microphone 110 is not passed back to the far-end telephone user. However, in the system of Figure 2, operation of the echo suppressor 130 is also made dependent upon the binary output 215 of the voice activity detector 215. More specifically, the echo suppressor 130 is activated only when the voice activity detector 210 indicates that the far-end signal 125 includes a speech component (e.g., when the detector output 215 is equal to binary one).

By conditioning activation of the echo suppressor 130 on the voice activity detector output 215, the system 200 of Figure 2 is less likely to introduce undesirable distortion on the microphone signal 115 when it is not necessary to do so. In other words, since there can be no far-end echo during periods in which the far-end user is not speaking, there is no need for the echo suppressor 130 to be active during those periods. Thus, by deactivating the echo suppressor 130 when the voice activity detector 210 indicates no far-end speech, the system 200 of Figure 2 reduces the likelihood that the echo suppressor 130 will attenuate, and thus distort, near-end speech and noise during periods when far-end echo suppression is not needed.

Note, however, that since the loudspeaker voice activity detector 210 can sometimes make mistakes, it cannot be guaranteed that the echo suppressor 130 will never be active when the far-end user is not speaking. Nonetheless, as compared to prior art systems which do not condition suppressor operation, the system 200 of Figure 2 provides superior near-end signal quality. Further, the system of Figure 2 provides superior protection against unnecessary attenuation by the echo suppressor 130 as compared to prior art systems which condition suppressor operation based on a microphone voice activity detector. Since the loudspeaker signal typically has a quieter background than the microphone signal (e.g., in an automobile hands-free application), the loudspeaker voice activity detector 210 is generally more reliable than is a microphone voice activity detector. In other words, it is easier to detect a lack of speech in a quiet-background loudspeaker signal than it is to detect low-level speech in a noisy-background microphone signal.

Note also that the above discussion presumes that the network equipment between the near-end and far-end users is sufficient to prevent the howling which can occur at frequencies for which the round-trip gain (i.e., the signal gain from the far-end user, across the network and back to the far-end user) is greater than unity. In other words, if the echo suppressor 130 is to be completely deactivated based on the absence of far-end speech, something else (e.g., a network echo canceler) should be used to prevent howling. Alternatively, instead of completely deactivating the echo suppressor 130 during periods of no far-end speech, the echo suppressor 130 can remain active and can be adjusted (in the case of clipping, scaling and hybrid suppressors) to provide a nominal level of attenuation which is just sufficient to prevent howling (e.g., -6 to -10 dB).

Those skilled in the art will appreciate that methods for implementing the loudspeaker voice activity detector 210 are well known. For example, European Telecommunications Standards Institute (ETSI) document GSM-06.82 describes an implementation suitable for purposes of the present invention. Such voice-activity detectors are currently used in many communications systems for purposes of noise suppression and/or discontinuous transmission (DTX). Typically, such voice activity

detectors are biased toward speech. In other words, they will sometimes indicate that speech is present when it is not present, but they will rarely indicate that no speech is present when it is present. Since the loudspeaker voice activity detector 210 of the present invention has similar requirements (i.e., it is preferable to err on the side of caution and attempt to attenuate echo that does not exist than it is to allow echo that does exist to be passed back to the far-end user), the same type of voice activity detector can be used in the exemplary system 200 of Figure 2.

As described above with respect to Figures 1A and 1B, the echo suppressor 130 of Figure 2 can be a clipping suppressor, a scaling suppressor, a hybrid suppressor or, when the front-end echo canceler 140 is included, a cut-off switch. Thus, depending upon which type of suppressor is used, any one of a variety of methods for deactivating the echo suppressor 130 during periods of no far-end speech can be implemented. For example, any type of suppressor can be effectively deactivated by using a bypass switch to circumvent the suppressor entirely. Alternately, the cut-off switch suppressor can be deactivated simply by forcing the switch closed. A clipping suppressor can be deactivated, or put in a pass-through mode, by setting the size of the clipping window to zero. A scaling suppressor, on the other hand, can be put in pass-through by setting the attenuation factor to unity. Furthermore, a hybrid suppressor can be put in pass through by setting the attenuation window to zero and/or setting the attenuation factor to unity.

Note that the embodiment of Figure 2 works especially well if the front-end echo canceler 140 of Figure 1B is included. This results from the fact that the echo canceler 140 can usually provide sufficient echo suppression, without assistance from the echo suppressor 130, during periods when the voice activity detector 210 incorrectly indicates that no speech is present in the loudspeaker signal 125. In other words, the voice activity detector 210 is likely to make such an error only when the loudspeaker signal 125 is relatively noisy (and the far-end speech is thus somewhat masked in noise), and the far-end user is less sensitive to noise-masked echo in the first instance. Whether the echo canceler 140 is included or not, however, the system 200 of Figure 2 provides excellent near-end signal quality.

Figure 3 depicts an alternative echo suppression system 300 constructed in accordance with additional aspects of the invention. As shown, the exemplary system 300 includes the microphone 110, the loudspeaker 120, the echo suppressor 130, the echo canceler 140 and the summing device 150 of Figures 1A and 1B, and a gain control processor 310 and a volume gain block 320. The microphone output 115 is coupled to the audio input of the acoustic echo canceler 140 and to a first reference input of the gain control processor 310. The audio output 145 of the acoustic echo canceler 140 is coupled to the audio input of the echo suppressor 130 and to a second reference input of the gain control processor 310. The suppressor audio output 135 and the comfort noise signal are coupled to the first and second inputs of the summing device 150, respectively, and the audio output 155 of the summing device 150 serves as the near-end audio input to the telephone (not shown).

The far-end audio output 125 from the telephone is coupled to an audio input of the volume gain block 320, and an audio output 325 from the volume gain block 320 is coupled to the audio input of the loudspeaker 120 and to reference inputs of the echo suppressor 130 and the acoustic echo canceler 140. The audio output 325 from the volume gain block 320 is also coupled to a third reference input of the gain control processor 310, and a control output 315 of the gain control processor 310 is coupled to a control input of the echo suppressor 130. A volume control signal 324 is coupled to a fourth reference input of the gain control processor 310 and to a control input of the volume gain block 320.

Generally, operation of the system 300 of Figure 3 is similar to that of the system 101 of Figure 1B. In other words, the acoustic echo canceler 140 dynamically models the acoustic path from the loudspeaker 120 to the microphone 110 and attempts to cancel any loudspeaker sound picked up by the microphone 110. The echo suppressor 130 then provides additional echo attenuation as necessary, and comfort noise is optionally added via the summing device 150 to compensate for near-end noise attenuated by the echo suppressor 130. In the system 300 of Figure 3, however, the gain control processor 310 controls the level of attenuation provided by the echo suppressor 130 in accordance with aspects of the invention.

The volume gain block 320 is included in the system 300 of Figure 3 to better illustrate operation of the gain control processor 310. Generally, the volume gain block 320 attenuates the telephone audio signal 125, in accordance with the volume control signal 324, and passes the attenuated audio signal 325 to the loudspeaker 120.

5 The volume control signal 324 can be adjusted, for example, by the near-end user.

According to one aspect of the invention, the echo suppressor 130 is presumed to be either a scaling suppressor or a hybrid suppressor which multiplies at least a portion of the audio signal 145 by an attenuation factor α . In such case, the gain control processor 310 utilizes the volume control signal 324, the microphone signal

10 115, the loudspeaker signal 125 and/or the canceler output signal 145 to provide the attenuation factor α via the control processor output signal 315.

Specifically, the gain control processor 310 utilizes the reference signals 115, 145, 324, 325 to dynamically measure a loop, or round-trip, echo gain which is in turn used to compute the suppressor attenuation factor α . The loop gain indicates the

15 degree to which far-end speech in the telephone audio signal 125 is being attenuated as it passes through the volume control block 320, across the acoustic path from the loudspeaker 120 to the microphone 110 and through the echo canceler 140. The loop gain can thus be used to compute the amount of additional attenuation the echo suppressor 130 should provide so that the echo in the near-end audio signal 135 is

20 attenuated to a particular goal level.

Note that the level of attenuation provided by the volume gain block 320 is known from the volume control signal 324. However, the level of attenuation provided by the acoustic path (which is referred to as the channel gain) and the level of

25 attenuation provided by the echo canceler 140 (referred to as the canceler gain) are computed based on the microphone signal 115, the loudspeaker signal 325 and the canceler output signal 145. Advanced methods for computing the channel and canceler gains are described, for example, in the above incorporated U.S. Patent Application No.09/005,149 (entitled "Methods and Apparatus for Improved Echo Suppression in Communications Systems" and filed on even date herewith). Since specific details

30 relating to computation of the channel and canceler gains are not critical to an

understanding of the present invention, however, a detailed description of such computation is omitted here for sake of brevity.

As noted above, the gain control processor 310 can, given the loop gain from the input of the volume control block 320 to the output of the echo canceler 140 (or to the output of the microphone 115 if the echo canceler 140 is not included), compute the suppressor attenuation factor α so that the echo in the near-end audio signal 155 is suppressed to a particular goal level. However, by way of contrast to prior art systems in which the goal level is fixed, the gain control processor 310 of the system 300 of Figure 3 adjusts the goal level based on prevailing conditions so that the echo suppressor 130 introduces a minimum of distortion on the near-end audio signal 155.

In other words, whereas prior art systems typically set the goal level to a relatively stringent fixed value (e.g., to account for a single worst case condition such as far-end single talk with no near-end background noise) and thus distort the near-end signal 155 more than is necessary when the stringent goal level is not required (e.g., when the single worst case condition does not exist), the gain control processor 310 of Figure 3 dynamically selects a preferred one of multiple possible goal levels, each one of the possible goal levels being most suitable for a particular prevailing condition. For example, during periods of far-end single talk, the far-end user is quite sensitive to echo, and the goal level should be commensurately severe (e.g., -45 dB or lower). During periods of double talk, however, the far-end user is less sensitive to echo, and the goal level can be somewhat relaxed (e.g., -25 dB or lower). Advantageously, the gain control processor 310 can detect these and other prevailing conditions and set the goal level (and thus the suppressor attenuation factor α) accordingly. As a result, the system 300 of Figure 3 introduces no more distortion on the near-end signal 155 than is necessary at any given time.

To detect conditions of near-end single talk, far-end single talk, double talk, high near-end noise, etc., the gain control processor 310 utilizes the microphone signal 115, the loudspeaker signal 325 and/or the canceler output signal 145. For example, to detect far-end speech, the gain control processor 310 can include the loudspeaker voice activity detector 210 described above with reference to Figure 2. Additionally, the

gain control processor 310 can include a microphone voice activity detector to detect near-end speech based on the microphone signal 115 and/or the canceler output signal 145.

Those skilled in the art will appreciate that methods for implementing a crude microphone voice activity detector based only on the microphone signal 115 are well known. Additionally, advanced methods for detecting near-end speech based on the input and output signals of an echo canceler (e.g., based on the microphone signal 115 and the canceler output 145) are described by way of pseudo-code below. Since specific details relating to detection of near-end speech are not critical to an understanding of the present invention, however, such details are omitted here.

It is important to note that the above described goal level selection process is applicable when the echo suppressor 130 is active and attenuating the audio signal 145 to at least some degree (i.e., when the echo suppressor 130 is not being bypassed or put in a pass through mode as described above). Thus, the present invention is distinguishable from systems which avoid near-end signal distortion by completely deactivating the echo suppressor 130 based on the detection of near-end speech (e.g., during periods of double talk). Since such systems rely on the echo canceler 140 alone to provide sufficient echo suppression when the echo suppressor 130 is deactivated, such systems are often unable to meet the recommended goals of today's telecommunications standards (e.g., IS-55A recommends -45 dB of echo attenuation during far-end single talk and -25 dB of echo attenuation during double talk).

By way of contrast, the present invention permits the echo suppressor 130 to remain active during periods of double talk and can therefor easily meet the recommended goals at all times. Empirical studies have shown that using echo suppression goals of -56 dB and -25 dB during periods of far-end single talk and double talk, respectively, provide a good balance between the conflicting needs for maximal echo suppression and minimal near-end signal distortion. Those skilled in the art will appreciate that, although goal levels are explicitly provided for only two prevailing conditions (i.e., far-end single talk and double talk), many other prevailing conditions

(e.g., high and low levels of near-end and far-end noise) are contemplated and can be accounted for by providing additional possible goal levels.

According to another aspect of the invention, the gain control processor 310 of Figure 3 regulates the echo suppressor 130 based upon a measurement of the ratio of echo to noise in the suppressor input signal 145. Specifically, the gain control processor 310 dynamically measures the echo-to-noise ratio at the suppressor input 145 and provides the echo suppressor 130 with an appropriate control signal via the processor output signal 315.

In one exemplary embodiment, the gain control processor 310 deactivates the echo suppressor 130 when the measured echo-to-noise ratio is below a predetermined threshold. In other words, when the echo is effectively hidden in, or masked by, background noise, the echo is not perceptible to the far-end user, and the echo suppressor 130 need not introduce near-end signal distortion in an attempt to suppress the echo further. Thus, the echo suppressor 130 can be deactivated or adjusted to provide a nominal level of attenuation which is just sufficient to prevent howling.

Empirical studies have shown that an echo-to-noise ratio of -10 dB or lower provides sufficient noise masking from a far-end user perspective. However, the noise masking threshold for deactivating the echo canceler 130 can be set lower than -10 dB in practice to avoid mistakenly passing perceptible echo due to errors in computation of the echo-to-noise ratio. Empirical studies have shown that a noise masking threshold of -18 dB is effective for this purpose.

As described above, the echo suppressor 130 can be deactivated in a variety of ways. For example, any type of echo suppressor 130 can be effectively deactivated using a bypass switch, and a cut-off switch suppressor can be deactivated by forcing the cut-off switch closed. Additionally, a clipping suppressor can be deactivated by reducing the size of the clipping window to zero, and a scaling suppressor can be deactivated by setting the suppressor attenuation factor to unity. A hybrid suppressor can be deactivated by setting the clipping window to zero and/or by setting the suppressor attenuation factor to unity.

Where the echo suppressor 130 is either a scaling suppressor or a hybrid suppressor (i.e., where the echo suppressor multiplies at least a portion of the suppressor input signal by an attenuation factor α), the echo-to-noise measurement can be combined with the above described loop gain concept to provide additional benefits. For example, in one embodiment of the invention, the gain control processor 310 computes the loop gain from the input of the volume gain block 320 to the input of the echo suppressor 130 as described above and then computes the suppressor attenuation factor α needed to mask the echo in background noise (i.e., the attenuation factor α which will result in an audio output signal 155 having an echo-to-noise ratio of, for example, -18 dB). Note that, since the echo suppressor 130 attenuates the entire audio signal 145 (including echo, speech and noise), the present embodiment presumes that comfort noise is added to the suppressor output signal 135 in order to compensate for the background noise removed by the echo suppressor 130.

According to exemplary embodiments, the ratio of echo to noise in the audio input to the echo suppressor 130 is computed by measuring the total energy in the audio input and by estimating a level of noise energy in the audio input. Specifically, the noise energy estimate is subtracted from the total energy measurement, and the resulting difference (which indicates a level of echo energy) is divided by the noise energy estimate to provide the estimated echo-to-noise ratio. Those skilled in the art will appreciate that methods for measuring the energy in a signal and for estimating a noise level within a signal are well known. Additionally, advanced methods for computing a noise level estimate are described by way of pseudo-code below. Since specific details relating to noise level estimation are not critical to an understanding of the present invention, however, such details are omitted here.

Advantageously, all of the above described aspects of the present invention can be combined within a single echo suppression system to provide sophisticated echo suppressor control and optimal near-end signal quality. To demonstrate this point, Figure 4 depicts an exemplary algorithm 400 in which the attenuation factor α for the echo suppressor 130 is adjusted using multiple aspects of the present invention. Since the exemplary algorithm 400 utilizes the loop gain concept, the algorithm 400 is

primarily applicable to scaling suppressors and hybrid suppressors. Those skilled in the art will appreciate, however, that certain aspects of the exemplary algorithm 400 are applicable to all types of suppressors as described above.

As shown, the algorithm 400 begins at step S10 and a decision is made at step S20 as to whether the loudspeaker voice activity detector 210 indicates the presence of far-end speech. If not, then the echo suppressor 130 is put in a pass through mode and the process is completed at step S60. However, if the loudspeaker voice activity detector 210 does indicate far-end speech, then a decision is made at step S30 as to whether near-end speech has also been detected.

If near-end speech has not been detected, then an attenuation factor required to achieve a predefined far-end single talk goal is computed at step S35. However, if near-end speech has been detected, then an attenuation factor required to achieve a predefined double talk goal is computed at step S40. In either case, an attenuation factor required to achieve a predefined noise masking goal is computed at step S45. Thereafter, at step S60, the attenuation factor α is set to the least stringent of the attenuation factors computed based on noise masking (at step S45) and speech detection (at either step S40 or S35). The process is then completed at step S60.

An exemplary embodiment of the algorithm 400 of Figure 4 is provided by way of pseudo-code below. In the pseudo-code, a scaling or hybrid echo suppressor is assumed to operate using an attenuation or gain factor echoSuppressorGain.

```

    if farEndVoiceFlag == 0,
        echoSuppressorGain = 1
    else,
        if nearEndVoiceFlag == 1,
            loopEchoGainGoal = -25 dB
        else,
            loopEchoGainGoal = -56 dB
        end
        echoSuppressorGain = ...
            max(loopEchoGainGoal / loopEchoGain, ...
                echoToNoiseGoal / echoToNoise)
    end

```

Note that in the algorithm of Figure 4, the noise masking goal is computed (at step S45) whether or not near end speech has been detected (at step S30). In practice, however, empirical studies have shown that, when the front-end echo canceler 140 is included, the attenuation factor α computed based on the typical double talk goal (e.g., -25 dB) is always greater (and thus causes less distortion) than that computed based on the typical noise masking goal (e.g., -18 dB). As a result, DSP cycles can be saved in practice (when the front-end echo canceler 140 is included) by not bothering to compute the attenuation factor α for the noise masking goal when double talk has been detected.

Also note that the algorithm 400 of Figure 4 does not attempt to deactivate the echo suppressor 130 when the echo-to-noise ratio at the suppressor input is already at or below the noise masking goal. This results from the fact that empirical studies have shown that, for scaling and hybrid suppressors, it is better to leave the echo suppressor 130 active (with an attenuation factor α based on the noise masking goal) so that perceptible error is not mistakenly passed due to inaccurate echo and noise estimates (e.g., the estimates used to compute the echo-to-noise ratio itself).

Computer simulations have shown that an echo suppression system constructed in accordance with the teachings of the present invention (e.g., in accordance with the algorithm of Figure 4) can significantly reduce the distortion of near-end speech and background noise typically introduced by a non-linear echo suppressor. For example, deactivating the echo suppressor 130 when the loudspeaker voice activity detector 210 indicates no far-end speech results in perfect, non-distorted near-end speech and noise under far-end single talk conditions. This is true even when there is constant noise in the loudspeaker signal.

Additionally, adjusting the echo suppression goal for an active echo suppressor based on the detection of near-end speech results in greatly improved near-end speech and noise under double talk conditions (and perfect near-end speech and noise when the combined gain of the volume control, the acoustic echo path and the linear echo canceler is -25 dB or better, which is usually the case). Adjusting the echo suppression goal for an active echo suppressor also serves to improve near-end speech and noise

under near-end single talk conditions when the loudspeaker voice activity detector 210 mistakenly detects noise as speech.

Further, controlling an echo suppressor based on a measurement of echo to noise at the suppressor input improves near-end speech and noise under noisy car conditions. For example, during far-end single talk, attenuation of the near-end background noise is reduced. As a result, less comfort noise is needed to reassure the far-end user, and the character of the background noise perceived by the far-end user is better preserved. Also, during periods of double talk or near-end single talk, near-end speech and noise are improved even when both the loudspeaker voice activity detector 210 and the near-end speech detector fail. Thus, during near-end single talk with a great deal of near-end noise (e.g., road and traffic noise in an automobile hands-free application), the present invention provides three mechanisms to ensure reduced distortion of near-end speech and noise despite constant loudspeaker noise.

By way of example, another embodiment of the algorithm 400 of Figure 4, as integrated in an actual echo suppression system, is described in the pseudo-code provided below. The pseudo-code is written to simulate an exemplary echo suppression system as implemented using a 32-bit digital signal processor. Those skilled in the art will appreciate that such code is exemplary in nature and that equivalents can be implemented using a wide variety of hardware configurations. The exemplary echo suppression system includes a modified version of the improved echo canceler described in the above referenced U.S. Patent Application No. 08/852,729 and an AC-center attenuator as described in the above referenced U.S. Patent Application No. _____ (entitled "Methods and Apparatus for Improved Echo Suppression in Communications Systems" and filed on even date herewith).

```

25  % AEC and ANLP simulation script for MATLAB.

    % Before running this script, set the following variables:
    %   inFile = name of input file, left = far end, right = near end.
    %   outFile = name of output file, left = ANLP output, right = AEC output.
    % All files use the raw format of the DAT-Link.
30  % estnoise.m contains the function to estimate noise.

    % Glossary:
    % EC = Echo Canceler = linear echo suppresser

```

```

% AEC = Acoustic-Echo Canceler = loudspeaker-echo canceler
% NLP = Non-Linear Process = residual-echo suppresser = AC-center
attenuator
5 % ANLP = Acoustic Non-Linear Process
% VAD = Voice-Activity Detector

% Maximum positive value for fractional representation.
ONE = 32767/32768;

% Read file containing far-end and near-end signals.
fidIn = fopen(inFile, 'r');
10 if fidIn == -1
    error(['Error opening file ' inFile])
end
[LRmatrix, wordCount] = fread(fidIn, [2,inf], 'int16');
fclose(fidIn);

15 % The number of samples in the update integration period.
FRAME_SIZE = 160;
% Larger frame sizes give greater robustness to double-talk & near-end
noise
% which tend to integrate towards zero.
20 % Larger also improves ability to detect convergence because the correlated
% update grows proportionally with frame size whereas the uncorrelated
(noise)
% grows with the square-root of the frame size.
% Smaller improves reaction time to changes (echo path, single -> double
25 talk)
% and speeds up convergence.
% Smaller also improves ability to reject vowel sounds.
% 160 is used for ease of porting to a 160-samples-per-frame TDMA phone.
% The resulting 20 ms frame is nearly optimum for dividing speech into
30 % stationary-signal segments.

NFRAMES = floor((wordCount/2)/FRAME_SIZE); % Number of frames to process.
clear wordCount
NSAMPLES = NFRAMES * FRAME_SIZE; % Length of sample-based vectors for
debug.

35 %
% Load the mic (uplink) and speaker (downlink) VAD outputs from separate
% files. If each file is not found, run the C executable on the near-end
and
% far-end sound files, and save the VAD outputs in a file with the same
40 % prefix.
%
fidIn = fopen([inFile '_up_vad'], 'r');
if fidIn == -1
    disp(['File = ' inFile '_up_vad not found -- creating ...'])
45 fidOut = fopen('vad_in.raw', 'w');
fwrite(fidOut, LRmatrix(2,:), 'int16'); % Uplink audio
fclose(fidOut);

```

```

!nrsim -s=f vad_in.raw junk vad_out.bit
!rm vad_in.raw
!rm junk.flt

5   fidIn = fopen('vad_out.bit','r');
    if fidIn == -1
        error('Error opening file = vad_out.bit')
    end
    micVad = fread(fidIn, NFRAMES, 'int8');
    !rm vad_out.bit
10  fclose(fidIn);

    fidOut =fopen([inFile '_up_vad'],'w');
    fwrite(fidOut, micVad, 'int8');
    fclose(fidOut);
else
15  micVad = fread(fidIn, NFRAMES, 'int8');
    fclose(fidIn);
end

fidIn = fopen([inFile '_down_vad'],'r');
if fidIn == -1
20  disp(['File = ' inFile '_down_vad not found -- creating ...'])
    fidOut =fopen('vad_in.raw','w');
    fwrite(fidOut, LRmatrix(1,:), 'int16');    % Downlink audio
    fclose(fidOut);

    !nrsim -s=f vad_in.raw junk vad_out.bit
25  !rm vad_in.raw
    !rm junk.flt

    fidIn = fopen('vad_out.bit','r');
    if fidIn == -1
        error('Error opening file = vad_out.bit')
    end
30  speakerVad = fread(fidIn, NFRAMES, 'int8');
    !rm vad_out.bit
    fclose(fidIn);

    fidOut =fopen([inFile '_down_vad'],'w');
    fwrite(fidOut, speakerVad, 'int8');
    fclose(fidOut);
35  else
    speakerVad = fread(fidIn, NFRAMES, 'int8');
    fclose(fidIn);
40  end

% Scale inputs to use range of -1 to ONE.
LRmatrix = LRmatrix / 32768;

% Number of bits to right shift values accumulated over a frame of samples.
FRAME_BITS = ceil(log2(FRAME_SIZE));

```

```

% Scale factor to quantize energies to 32 bits (Z8.23 format
w/FRAME_SIZE=160)
ENERGY_SCALE = 2^(31-FRAME_BITS);

% Number of taps in the reference delay line.
5 % It must be long enough that the high-delay taps are mostly uncorrelated
with
% the far-end signal and only have near-end energy.
AEC_REF_TAPS = 512;

% Number of taps in the FIR echo-estimation filter.
10 AEC_COEF_TAPS = 256;

% Number of taps in the high-delay section of the update vector for
measuring
% near-end energy.
AEC_NEAR_TAPS = 128;

15 % Length of vector for capturing car noise samples.
COMFORT_NOISE_SIZE = 128;

% Calculate the threshold on the update vector peak-to-baseline ratio for
% determining the maximum update gain. For noise uniform in the range of
% [-1 1], the expected peak update magnitude is g*FRAME_SIZE/3, while the
20 % expected RMS of the update vector for the uncorrelated taps is
%  $g \cdot \sqrt{\text{FRAME\_SIZE}}/3$ , where g is the echo path gain. Therefore, the
maximum
% update vector peak-to-baseline ratio is  $\sqrt{\text{FRAME\_SIZE}}$ .
% However, voice does not have a constant envelope like noise. Because
25 % FRAME_SIZE is much smaller than the update vector length, a burst of
speech
% will sometimes be in the area of the update vector where the peak is
% measured but not in the area where baseline is measured. Therefore, real
% peak-to-baseline ratios can be extremely high (>300).
30 % Setting the threshold too low will cause instability due to high-gain
% updates even for low-energy far-end signals under noisy or double-talk
% conditions.
% Setting the threshold too high will cause slow adaptation due to
high-gain
35 % updates only for high-energy far-end signals after large echo-path
changes.
% This threshold was empirically determined as a compromise.
AEC_MAX_GAIN_THRESH = 16;

% The peak-to-RMS for noise is independent of the frame size. This
40 threshold
% for rejecting near-end voice/noise and far-end periodic signals (tones
and
% vowels) was empirically derived.
AEC_BASELINE_THRESH = 5.5;

45 % Pre-calculate the constant to use as a multiplier for the status gauge.
AEC_STATUS_GAUGE_SCALER = floor(32768/...
```



```

                                (AEC_MAX_GAIN_THRESH-AEC_BASELINE_THRESH)) /
32768;

% Create the gain profile for the FIR coefs. The profile roughly matches
the
5 % expected range of the coefs in the car. This way, updates containing
% periodic components (vowels) are forced to follow the proper exponential
% decay characteristic and minimize divergence. Lower gain on the higher-
% delay taps also reduces update noise contribution to the coefs. The
overall
10 % effect of the profile is to allow higher update gain without instability.

% Since all coefs have 16-bits of dynamic range, the higher-delay taps also
% have better quantization as a result of the profile. The profile is
% implemented by calculating the FIR 64 taps at a time with a right shift
in
15 % between.
profile=ones(AEC_COEF_TAPS,1);
for k=2:(AEC_COEF_TAPS/64),
    profile((k*64-63):(k*64))=ones(64,1)*2^(1-k);
end

20 % Allocate debug vectors to speed up execution.
aecUpdateFactor = zeros(1,NFRAMES);
aecChanGainHist = zeros(1,NFRAMES);
aecSpeedHist = zeros(1,NFRAMES);
aecVoiceGainHist = zeros(1,NFRAMES);
25 aecVoiceGainBaseHist = zeros(1,NFRAMES);
aecNearRatioHist = zeros(1,NFRAMES);
aecNearGainHist = zeros(1,NFRAMES);
aecEchoGainHist = zeros(1,NFRAMES);
aecInNoiseHist = zeros(1,NFRAMES);
30 aecInEchoNoiseHist = zeros(1,NFRAMES);
aecInVoiceHist = zeros(1,NFRAMES);
aecInEchoVoiceHist = zeros(1,NFRAMES);
anlpInVoiceHist = zeros(1,NFRAMES);
anlpInNoiseHist = zeros(1,NFRAMES);
35 anlpDeltaHist = zeros(1, NSAMPLES);
anlpGainHist = zeros(1,NFRAMES);

% Initialize variables
aecRef = zeros(1,AEC_REF_TAPS);
values: % To use the last
40 aecCoef = zeros(AEC_COEF_TAPS,1); % Comment out
this
aecInNoise = FRAME_SIZE; % Comment out
this
45 aecInEchoNoise = aecInNoise; % Comment out
this
anlpInNoise = aecInNoise; % Comment out
this
aecChanGain = ONE; % Comment out
this

```

```

aecVoiceGain = ONE;                                % Comment out
this
aecVoiceGainBase = aecVoiceGain;                    % Comment out
this
5  aecNearGain = aecVoiceGain;                        % Comment out
this
aecEchoGain = aecVoiceGain;                          % Comment out
this
10 anlpComfortNoiseInOld = 0;                          % Comment out
this
anlpComfortNoiseOutOld = 0;                          % Comment out
this
anlpComfortNoise = zeros(1,COMFORT_NOISE_SIZE);      % Comment out
this
15 anlpArCoef = 0.75;                                  % Comment out
this
aecNearRatio = 0;                                     % Init for history only
aecInNoiseStateVars = [aecInNoise 0 0];
aecInEchoNoiseStateVars = [aecInEchoNoise 0 0];
20 anlpInNoiseStateVars = [anlpInNoise 0 0];
anlpSeed = 1;
anlpArGain = 1 - anlpArCoef;
anlpRefEnvelope = 0;
anlpOutLast = 0;
25 anlpNearSpeechCount = 0;
anlpNearSpeechFlag = 0;

%disp(['aecCoef(1) = ' dec2hex(aecCoef(1)*32768+(aecCoef(1)<0)*65536)])
%disp(['aecCoef(2) = ' dec2hex(aecCoef(2)*32768+(aecCoef(2)<0)*65536)])
%disp(['aecInNoise = ' dec2hex(aecInNoise*2^31)])
30 %disp(['anlpInNoise = ' dec2hex(anlpInNoise*2^31)])
%disp(['aecChanGain = ' dec2hex(aecChanGain*32768)])
%disp(['aecVoiceGain = ' dec2hex(aecVoiceGain*32768)])
%disp(['aecVoiceGainBase = ' dec2hex(aecVoiceGainBase*32768)])
%disp(['aecEchoGain = ' dec2hex(aecEchoGain*32768)])
35 %disp(['anlpComfortNoiseInOld = ' dec2hex(anlpComfortNoiseInOld*32768)])
%disp(['anlpComfortNoiseOutOld = ' dec2hex(anlpComfortNoiseOutOld*32768)])
%disp(['anlpArCoef = ' dec2hex(anlpArCoef*2^31)])
%disp(['anlpArGain = ' dec2hex(anlpArGain*32768)])

fidOut = fopen(outFile, 'w');

40 for frame = 1:NFRAMES,
    frame                                % Display the frame number to indicate
    progress.

    %
    % AEC pre-frame section
45 %

    % Since there is a gap between taps of the reference vector which are
    used

```

```

    % to update the FIR coefficients and those used in correlation of
near-end
    % energy, the update vector need not be calculated for every tap of the
    % reference vector. Therefore, the update vector is represented by sub-
5   % vectors specifically for the two purposes.
    % Clear update sub-vectors which accumulate over a frame.
    aecUpdate = zeros(AEC_COEF_TAPS,1);           % Used for FIR coef update
    aecUpdateNear = zeros(AEC_NEAR_TAPS,1);       % Used for near-end
measurement
10   % Clear other frame accumulators
    aecEchoEstEnergy = 0;

    % Reset block-floating-point variables.
    aecShiftPending = 0;
    aecErrorShift = 0;

15   % Get uplink and downlink PCM audio samples into buffers.
    downlinkAudio = LRmatrix(1, (frame-1)*FRAME_SIZE+1 : frame*FRAME_SIZE);
    uplinkAudio = LRmatrix(2, (frame-1)*FRAME_SIZE+1 : frame*FRAME_SIZE);

    % Accumulate AEC near-end-input energy over a frame.
    aecInEnergy = sum(uplinkAudio.^2);
20   % Quantize energy to 32 bits.
    aecInEnergy = floor(aecInEnergy * ENERGY_SCALE) / ENERGY_SCALE;

    %
    % AEC sample section
    %

25   for k = 1:FRAME_SIZE,
        % Shift the far-end (loudspeaker) sample into the reference delay line
and
        % calculate FIR output.
        % In the DSP, both operations are in one instruction.
30         aecRef = [downlinkAudio(k) aecRef(1:AEC_REF_TAPS-1)];
%TEST CODE START
%The following code quickly approximates the commented-out, bit-accurate
code.
35         aecEchoEst = aecRef(1:AEC_COEF_TAPS) * (aecCoef .* profile);
        aecEchoEst = max(min(round(aecEchoEst * 32768)/32768,ONE),-1);
%TEST CODE END
        % aecEchoEst = 0;
        % for m=(AEC_COEF_TAPS/64):-1:2,
        % aecEchoEst = aecEchoEst + aecRef(m*64-63:m*64) *
40         aecCoef(m*64-63:m*64);
        % aecEchoEst = max(min(aecEchoEst,ONE),-1) / 2;
        % Quantize for S.31 format
        % aecEchoEst = floor(aecEchoEst * 2^31) / 2^31;
        % end
45         % aecEchoEst = aecEchoEst + aecRef(1:64) * aecCoef(1:64);
        % aecEchoEst = max(min(aecEchoEst,ONE),-1);
        % Quantize for S.15 format

```

```

%      % Add 2^(-17) to force the 1's complement floating point to act the
same
%      % as 2's complement when rounding a negative number with a fraction of
%      % exactly 0.5.
5      %      aecEchoEst = round(aecEchoEst * 32768 + 2^(-17))/32768;
%      aecEchoEst = max(min(aecEchoEst,ONE),-1);

% Accumulate echo-estimate energy over a frame.
% To improve small-signal performance and to make this measurement in
the
10      % same way as the other energy accumulations, the full 40-bit
accumulator
% is saved between loop passes.
aecEchoEstEnergy = aecEchoEstEnergy + aecEchoEst^2;

% Calculate the AEC output = near-end (microphone) input - echo
15      estimate.
uplinkAudio(k) = max(min(uplinkAudio(k) - aecEchoEst,ONE),-1);
% Accumulate coef update = correlation of error (uplinkAudio(k)) and
% reference. Use block floating point representation, where
aecErrorShift
20      % is the exponent and aecUpdate/aecUpdateNear() is the mantissa.
T = uplinkAudio(k) * 2^aecErrorShift;
% Quantize for S.15 format
T = floor(T * 32768)/32768;
if aecShiftPending,
25      ASM = -1;
      aecErrorShift = aecErrorShift - 1;
      aecShiftPending = 0;
else
30      ASM = 0;
end
% Calculate for the region used to update the FIR coefficients.
aecUpdate = aecUpdate + T * aecRef(1:AEC_COEF_TAPS)';
% Calculate for the region used to measure near-end energy.
aecUpdateNear = aecUpdateNear + ...
35      T * aecRef(AEC_REF_TAPS-AEC_NEAR_TAPS+1:AEC_REF_TAPS)';
% Quantize for S.15 format
% Add 2^(-17) to force the 1's complement floating point to act the
same
40      % as 2's complement when rounding a negative number with a fraction of
% exactly 0.5.
aecUpdate = round(aecUpdate * 32768 + 2^(-17))/32768;
aecUpdateNear = round(aecUpdateNear * 32768 + 2^(-17))/32768;
aecUpdate = max(min(aecUpdate,ONE),-1);
45      aecUpdateNear = max(min(aecUpdateNear,ONE),-1);
aecUpdate = aecUpdate * 2^ASM;
aecUpdateNear = aecUpdateNear * 2^ASM;
% Quantize for S.15 format after possible right shift.
aecUpdate = floor(aecUpdate * 32768)/32768;
50      aecUpdateNear = floor(aecUpdateNear * 32768)/32768;
% Find the peak square of the update vector (assume in first 128 taps).
% The goal is to get the peak absolute value, but the peak square takes

```

```

    % fewer cycles in the DSP, even with the sqrt at the end of the frame.
    aecUpdatePeak2 = max(aecUpdate(1:128).^2);
    % Flag indicates if update needs divided by 2 in the next loop.
    aecShiftPending = aecUpdatePeak2 > 0.25;    % 0.25 = 0.5^2
5   end

    % Quantize energy for 32-bits.
    aecEchoEstEnergy = floor(aecEchoEstEnergy * ENERGY_SCALE) / ENERGY_SCALE;

    %
10  % AEC post-frame section
    %

    aecOut = uplinkAudio;    % Save for output to file for debug.

    % Accumulate AEC-output energy over a frame.
    aecOutEnergy = sum(uplinkAudio.^2);
    % Quantize energy for 32-bits.
15  aecOutEnergy = floor(aecOutEnergy * ENERGY_SCALE) / ENERGY_SCALE;

    % The true reference energy is different for each element of the update
    % vector. aecUpdate(1) would use the energy from aecRef(1),
    % aecUpdate(2) would use the energy from aecRef(2), and so forth. To
    % reduce complexity, use a single number to represent the reference
20  energy.
    % When the reference energy is used to measure the channel echo gain
    % for determining adaptation speed or to normalize the update (NLMS),
    using
    % too small of a value could cause instability. The compromise solution
25  % implemented here is to use the maximum of the endpoints where profile =
    1.
    aecRefEnergy = max(sum(aecRef( 1:FRAME_SIZE ) .^2), ...
                       sum(aecRef(64:FRAME_SIZE+63) .^2));
    % Quantize for Z8.7 format
30  aecRefEnergy = floor(aecRefEnergy*128)/128;

    % Measure the update baseline as the RMS of the high-delay elements where
    % the correlation between the error and far-end signals is expected to be
    0.
    % Add 1 LSB to the result to ensure aecUpdateBase is greater and
35  % aecPeakToBase is smaller after quantization. This avoids the false
    % impression of higher echo correlation.
    % Adding 1 LSB after a floor operation produces the same result as a
    ceiling
    % operation except for the rare case when all the truncated bits equal
40  zero.
    aecUpdateBase = sum(aecUpdateNear.^2)/AEC_NEAR_TAPS;
    % Quantize squared intermediate result for Z.31 format
    aecUpdateBase = floor(aecUpdateBase*2^31)/2^31;
    aecUpdateBase = sqrt(aecUpdateBase);
45  % Quantize for Z.15 format
    aecUpdateBase = floor(aecUpdateBase*32768+1)/32768;

```

```

% Find the peak magnitude of the update vector.
aecUpdatePeak = sqrt(aecUpdatePeak2);

% Calculate the update peak-to-baseline ratio.
aecPeakToBase = aecUpdatePeak / aecUpdateBase;
5 % Quantize for Z11.4 format since 4 fractional bits are sufficient.
aecPeakToBase = floor(aecPeakToBase*16)/16;

% Calculate the status gauge (range=[0,ONE]) from the update peak-to-
% baseline ratio. The gauge, used in down-stream processing, stays the
same
10 % even though the peak-to-baseline ratio changes with frame size and the
% baseline threshold could change.
% For near-end voice/noise or far-end periodic signals (vowels), gauge <
0.1
% For example, the first frame of a DTMF tone, with frequencies of 941
15 Hz
% and 1209 Hz, was found to produce aecPeakToBase=3.5 and, thus,
% aecStatusGauge=0).
% For double talk, gauge < 0.3.
% For far-end single talk:
20 % gauge = ONE when canceler is grossly unconverged, regardless of
noise.
% If the near-end is quiet, gauge=ONE until near complete convergence.
% As the canceler converges, only residual echo higher in energy than
the
25 % near-end noise level causes gauge=ONE.
% Thus, near-end noise causes fewer updates, not reduced update gain.
aecStatusGauge = (aecPeakToBase - AEC_BASELINE_THRESH) * ...
AEC_STATUS_GAUGE_SCALER;
aecStatusGauge = max(min(aecStatusGauge,ONE),0);
30 % Quantize for Z.15 format
aecStatusGauge = floor(aecStatusGauge*32768)/32768

% Estimate the noise frame energy at the AEC input.
[confirmedNoVoiceFlag, aecInNoiseStateVars] = estnoise(aecInEnergy, ...
(micVad(frame)==0) & (aecStatusGauge==0), aecInNoiseStateVars);
35 aecInNoise = aecInNoiseStateVars(1);
aecInNoise % Display for status
aecInNoiseHist(frame) = aecInNoise; % Save for debug

% Estimate the echo gain at the AEC input (channel gain).
% Update the estimate only during far-end single talk
40 % (speakerVad(frame) == 1) & (aecStatusGauge > 0.3), when the ratio is
% accurate despite that the reference measurement includes voice and
noise
% while the AEC input measurement includes only voice (aecInVoice >
% aecInNoise*8), and when reference energy is not significantly affected
45 by
% quantization (aecRefEnergy >= 10/128). The instantaneous gain
measurement
% is not very accurate because of time misalignment and spectrum
variations.

```

```

% Therefore, an averaging process is used. The norm of the echo canceler
% filter coefficients is an excellent long-term estimate of the channel
% gain. However, it does not track a changing echo path quickly enough.
% Also, since the channel gain estimate controls the adaptation speed of
5 the
% linear echo canceler, stability is improved if the estimate is
independent
% of the linear echo canceler as much as possible. Thus, the following
% scheme is used:
10 % If the instantaneous measurement (aecChanGainTrial) is less than the
% estimate (aecChanGain), pump the estimate down exponentially.
% To speed adaptation, the estimate is set directly to the instantaneous
% measurement if the error is greater than 50%, and the estimate is
pumped
15 % down proportionally if the error is greater than 12.5%, i.e.,
% (aecChanGain-aecChanGainTrial)/4 > aecChanGain/32 for
% aecChanGain-aecChanGainTrial > aecChanGain/8 = aecChanGain*0.125.
% If the instantaneous measurement is greater than the estimate, pump the
% estimate up exponentially only if the measurement is not clearly
20 dominated
% by near-end voice (aecChanGainTrial < 2). It would not work to use a
% relative comparison such as (aecChanGainTrial < 2*aecChanGain) because
% no update would occur when the true channel gain jumps quickly.
% Using a pump-up time constant that is 1/8 of the pump-down time
25 constant
% helps tolerate near-end voice in the instantaneous measurement.
% It is difficult to tell the difference between near-end speech and when
% channel echo gain gets worse. The pump-up time constant here
% determines the tracking rate, and it was empirically determined.
30 aecInVoice = max(0, aecInEnergy-aecInNoise);
aecInVoiceHist(frame) = aecInVoice;
if (speakerVad(frame) == 1) & (aecStatusGauge > 0.3) & ...
    (aecInVoice > aecInNoise*8) & (aecRefEnergy >= 10/128),
35 % Quantize aecInVoice to 428.20 format for use as dividend to get the
% desired scale for the quotient.
aecChanGainTrial = floor(aecInVoice *2^20)/2^20 / aecRefEnergy;
% Quantize and limit quotient to Z2.13 format
aecChanGainTrial = min(4*ONE, floor(aecChanGainTrial *2^13)/2^13);
% Calculate the square root of the quotient.
40 aecChanGainTrial = sqrt(aecChanGainTrial);
% Quantize root to Z1.14 format
aecChanGainTrial = floor(aecChanGainTrial *2^14)/2^14;
if (aecChanGainTrial < aecChanGain/2),
    aecChanGain = aecChanGainTrial;
45 elseif (aecChanGain > aecChanGainTrial),
    aecChanGain = aecChanGain - ...
        max((aecChanGain-aecChanGainTrial)/4, aecChanGain/32);
elseif (aecChanGainTrial < 2*ONE),
    aecChanGain = min(ONE, aecChanGain + max(aecChanGain/256, 2^(-11)));
50 end
% Quantize for Z.15 format
aecChanGain = floor(aecChanGain*32768)/32768;
% Save for debug

```

```

    aecChanGainHist(frame) = aecChanGainTrial;
    else
        aecChanGainHist(frame) = ONE;
    end
5   aecChanGain                % Display for status
    aecSpeedHist(frame) = aecChanGain;    % Save for debug

    % Determine the update gain.
    % Use NLMS to make the adaptation speed constant (independent of far-end
    % signal amplitude) as long as the gain is less than or equal to ONE.
10   % Using the max function results in faster convergence than adding 1 to
    % the denominator because the resulting gain is higher.
    % Using the maximum of the AEC reference and near-end-input energies
    limits
15   % the normalizing gain when there is near-end noise and/or voice. The
    AEC
    % output energy is not used instead of the AEC near-end-input energy
    because
    % higher gain is not desired after convergence, and because stability is
    % improved by not using another parameter based on the AEC output to
20   control
    % the AEC update gain. aecInEnergy should be less than aecRefEnergy with
    no
    % near-end voice or noise in order to avoid microphone overload (since
    the
25   % microphone gain is set according to the loudest near-end speech level).
    % The energy multiplier is set to 8 if we are sure there is far-end
    single
    % talk with low near-end noise (aecInEnergy/16 >= aecOutEnergy).
    Otherwise,
30   % it is set to 16. Thus, adaptation is faster when the car is quiet.
    % A too-small energy multiplier results in less stability, causing
    % overshoots in the adaptation and spikes in the echo canceler output.
    % The overshoots also hinder differentiation between far-end single talk,
    % with a changing echo path, and near-end speech.
35   % A too-large multiplier increases echo gain shortly after a
    perturbation.
    % In the numerator, use the gauge to vary the gain according to the
    % conditions. Also, use the channel echo gain as a multiplier to
    % optimize adaptation speed to the channel. Without this multiplier,
40   % adaptation is either slower than necessary for high channel
    % gain or unstable for low channel gain. Using the norm
    % of the adaptive filter coefficients instead of the energy-based channel
    % gain results in a more accurate and consistent estimate in the
    % long term, but speed and stability would be compromised in the short
45   term
    % after an echo path change.
    if (aecInEnergy/16 >= aecOutEnergy),
        aecDenom = max(1, 8 * max(aecRefEnergy, aecInEnergy));
        % For debug, set the factor equal to the negative gauge value.
50   aecUpdateFactor(frame) = -aecStatusGauge;
    else
        aecDenom = max(1, 16 * max(aecRefEnergy, aecInEnergy));

```



```

    % For debug, set the factor equal to the gauge value.
    aecUpdateFactor(frame) = aecStatusGauge;
end
5   % Quantize for Z12.3 format
    aecDenom = floor(8 * aecDenom)/8;
    aecNumer = aecChanGain * aecStatusGauge;
    % Quantize for 22Z.18 format
    aecNumer = floor(aecNumer * 2^18)/2^18;
10   aecUpdateGain = min(aecNumer / aecDenom, ONE);
    % Quantize for Z.15 format
    aecUpdateGain = floor(aecUpdateGain*32768)/32768;

    % Add the update vector to the coefficient vector using the adaptive
    gain.
15   % aecCoef is multiplied by profile before use as FIR coefficients.
    aecCoef = aecCoef + (aecUpdate * 2^(-aecErrorShift) * aecUpdateGain);
    % Quantize for S.15 format
    % Add 2^(-17) to force the 1's complement floating point to act the same
    % as 2's complement when rounding a negative number with a fraction of
    % exactly 0.5.
20   aecCoef = round(aecCoef * 32768 + 2^(-17))/32768;
    aecCoef = max(min(aecCoef,ONE),-1);

    % The noise suppresser would go here and process uplinkAudio.
    % By using aecUpdate and aecUpdateNear only within one subroutine, they
    can
25   % be in temporary memory, available for overlay.

    %
    % ANLP pre-frame section
    %

    % Accumulate energy at the ANLP input, which is connected through the
30   noise
    % suppresser to the AEC output.
    anlpInEnergy = sum(uplinkAudio.^2);
    % Quantize energy to 32 bits.
    anlpInEnergy = floor(anlpInEnergy * ENERGY_SCALE) / ENERGY_SCALE;

35   % Estimate the noise frame energy at the ANLP input.
    % Use speakerVad since the echo estimate comes from the loudspeaker
    signal.
    [confirmedNoVoiceFlag, anlpInNoiseStateVars] = estnoise(anlpInEnergy, ...
        (micVad(frame)==0) & (aecStatusGauge==0), anlpInNoiseStateVars);
40   anlpInNoise = anlpInNoiseStateVars(1);
    anlpInNoiseHist(frame) = anlpInNoise;           % Save for debug

    % Calculate the comfort noise when no voice is confirmed.
    if confirmedNoVoiceFlag,
45   anlpComfortNoise = uplinkAudio(1:COMFORT_NOISE_SIZE);
    % Use the NLMS algorithm to estimate anlpArCoef in the first-order
    % ARMA noise model of the form:
    %      (1 - anlpArCoef)*(1 + 0.8125*Z^-1)/(1 - anlpArCoef*Z^-1).

```

```

% This ARMA model will be used to filter white noise to get noise that
% sounds like the car noise.
% The NLMS algorithm tries to minimize the following expression:
%   error = uplinkAudio * (1 - anlpArCoef*Z^-1)/(1 + 0.8125*Z^-1).
5 % The correlation between error and uplinkAudio*Z^-1 is the update
% to the coefficient estimate.

% anlpInEnergy is used to normalize the update gain to provide an
% adaptation rate independent of level. Calculate the non-zero
% denominator outside the loop to save MIPS.
10 anlpDenom = 4 * max(1/128, floor(anlpInEnergy*128)/128);
error = 0;
for i=2:FRAME_SIZE,
    error = - 0.8125 * error;
    % Quantize for 5S4.31 format (maximum of 10 * uplinkAudio)
15 error = floor(error * 2^31)/2^31;
    error = error + uplinkAudio(i-1:i) * [-anlpArCoef; 1];
    quotient = uplinkAudio(i-1:i-1) * floor(error * 2^12)/2^12 /
anlpDenom;
    quotient = max(min(quotient,ONE),-1);
    % Quantize for S.15 format
    quotient = floor(quotient * 32768)/32768;
    anlpArCoef = anlpArCoef + quotient;
    anlpArCoef = max(min(anlpArCoef,ONE),-1);
end
25 anlpArGain = 1 - anlpArCoef;
% Plot the frequency response of the comfort noise for debug.
%   plot((0:8191)/8192*4000,...
%       20*log10(abs(freqz(anlpArGain*[1 0.8125],[1 -anlpArCoef],8192))))
%   axis([0 4000 -40 10])
30 %   pause(1)
end

% Estimate the noise frame energy of the echo at the AEC input.
% Use speakerVad since the echo estimate comes from the loudspeaker
signal.
35 [confirmedNoVoiceFlag, aecInEchoNoiseStateVars] = estnoise(...
    aecEchoEstEnergy, (speakerVad(frame)==0) & (aecStatusGauge==0), ...
    aecInEchoNoiseStateVars);
aecInEchoNoise = aecInEchoNoiseStateVars(1);
aecInEchoNoiseHist(frame) = aecInEchoNoise; % Save for debug
40 test1Hist(frame) = aecInEchoNoiseStateVars(2); % Save for debug
test2Hist(frame) = aecInEchoNoiseStateVars(3); % Save for debug
aecInEchoNoiseStateVars % Display for
status

% Estimate the voice energy estimates. Don't let them go below zero.
45 aecInEchoVoice = max(0, aecEchoEstEnergy - aecInEchoNoise);
aecInEchoVoiceHist(frame) = aecInEchoVoice;
aecNearVoice = max(0, aecInVoice - aecInEchoVoice);
anlpInVoice = max(0, anlpInEnergy - anlpInNoise);
anlpInVoiceHist(frame) = anlpInVoice;

```

```

% Leak upward the linear-echo-canceller baseline echo-gain estimate.
% It is used for comparison to the instantaneous echo gain to detect
% near-end speech and for the ANLP gains during near-end speech.
5 % It leaks upward fast enough to track when the instantaneous
% echo gain gets worse. The leakage was empirically determined as a
% compromise between fast tracking to avoid false detection of near-end
% speech during far-end single talk with a changed echo path and
minimizing
% distortion of near-end voice/noise during far-end voice/noise.
10 % When far-end single talk starts after the echo path has changed,
% aecNearGain can go down, go up somewhat, and then go way down. If
% aecNearGain goes up sufficiently above aecVoiceGainBase, near-end
speech
15 % will be detected. If this happens, the leakage on aecVoiceGainBase can
be
% increased to prevent this. A leakage constant of 5/4096 was
empirically
% found to provide sufficient tracking speed. However, increased leakage
% reduces detection of near-end speech during double talk.
20 aecVoiceGainBase = min(aecVoiceGainBase + 1/8192, ONE);

% Measure the linear-echo-canceller voice gain, excluding the channel.
% Including the channel gain would make near speech detection unreliable.
% Compute the root of the ratio of voice energy at the output and input
of
25 % the linear echo canceler.
% This raw measurement is not conditional as to whether the voice comes
from
% the near end, far end, or both.
% The output of the linear echo canceler is taken at the ANLP input
30 because
% the signal has passed through the noise suppresser, making measurements
% more accurate. Measurement by means of energy includes the loudspeaker
% distortion in the echo, which cross correlation does not. Also,
comparing
35 % energies at the input and output of the echo canceler avoids time-
% alignment issues that arise when comparing the echo-canceller output to
the
% reference (due to the delay of the channel).
% Measure gain only when not corrupted by quantization (aecInVoice >
40 8/2^7),
% when voice is present (aecInVoice > aecInNoise*8),
% and when residual echo is measurable (anlpInVoice > anlpInNoise/2).
% The accuracy of the measurement is not reduced by periodic components
% in the far-end signal.
45 if (aecInVoice > max(8/2^7, aecInNoise*8)) & ...
    (anlpInVoice > anlpInNoise/2),
    aecVoiceGain = min(ONE, sqrt(anlpInVoice /
(floor(aecInVoice*2^7)/2^7)));
% Quantize for Z.15 format
50 aecVoiceGain = floor(aecVoiceGain*32768)/32768;

% Measure the ratio of near-end voice to total voice.

```

```

% This produces fewer false indications of near-end voice due to a
% changing echo path with far-end single talk because it only relies on
% the energy of the echo estimate, not how well the echo is canceled at
% the output. However, like aecVoiceGain, false indications of
5 near-end
% voice are likely when the canceler is grossly untrained.
% The usual ratio would include a square root since the voice
measurements
% are in energy units. However, greater differentiation is achieved
10 % between near-end voice and poor canceler training by using the energy
% ratio directly. This also saves MIPS.
aecNearRatio = min(ONE, aecNearVoice / (floor(aecInVoice*2^7)/2^7));
% Quantize for Z.15 format
aecNearRatio = floor(aecNearRatio*32768)/32768;

15 % Measure the linear-echo-canceler baseline echo-gain estimate during
far-
% end speech. Measure whenever far-end single talk could exist
% (aecStatusGauge > 0.1) to make sure the parameter tracks the true
echo
20 % gain (not noise). Otherwise, echo may be heard. Update the
% baseline echo-gain estimate when the voice gain is lower than the
% baseline estimate because this indicates a high probability of
far-end
% single talk. By using leakage and not letting the baseline track the
25 % voice gain when it is higher than the baseline, the baseline is very
% robust against noise and near-end speech.
if (aecStatusGauge > 0.1),
    aecVoiceGainBase = min(aecVoiceGainBase, aecVoiceGain);
end

30 % Measure the linear-echo-canceler voice gain during near-end
% speech, and reset during far-end single talk (take minimum with voice
% gain). This will be used for comparison against the voice gain
% baseline to detect double talk. The only indicators available that
% near-end speech may be occurring are that the echo canceler gauge is
35 low
% (aecStatusGauge < 0.3) and that the echo canceler output contains
% sufficient energy that the voice is likely not just residual echo
% (anlpInVoice > anlpInNoise*8). Remember that the gauge could be low
due
40 % to periodic components in the far-end speech. Therefore, extra means
% are necessary to differentiate between when voice gain gets
% suddenly worse due to a change in the echo path or loudspeaker
% distortion and when there is near-end speech. When voice gain
% gets suddenly worse during far-end single talk, the duration tends to
45 be
% rather short. Therefore, filtering the increases in this parameter
% usually rejects far-end single-talk incidents. The filter time
constant
50 % is a compromise between far-end single-talk rejection and double-talk
% recognition speed, and it was empirically determined.
aecNearGainLast = aecNearGain;

```

```

aecNearGain = min(aecNearGain, aecVoiceGain);
if (aecStatusGauge < 0.3) & (anlpInVoice > anlpInNoise*8),
    aecNearGain = aecNearGain + (aecVoiceGain-aecNearGain)/8;
    % Quantize for Z.15 format
5    aecNearGain = floor(aecNearGain*32768)/32768;
    end

    % Use any of three detection methods for near-end speech. Each
    % method is more sensitive to near-end speech under certain conditions.
    % All detection occurs only when the total voice is sufficiently
10    greater
    % than an absolute level or the noise floor to reject conditions where
    % only noise is present.
    %
    % Method 1.
15    % (aecInEchoVoice < aecInVoice/4) is equivalent to (aecNearRatio >=
    0.75)
    % but the former is less sensitive to quantization. Spikes on
    % aecNearRatio are common for low levels of far-end single talk.
    % Therefore, the threshold must be high to avoid false detection. The
20    % only qualifiers needed for this detection method is that the status
    % gauge = 0 and the voice gain baseline be below 0.4625 = 15155/32768.
    % The latter threshold is an empirical compromise between avoiding
    false
    % detection when the canceler is grossly untrained and detecting double
25    % talk as soon as possible after the canceler starts training.
    %
    % Method 2.
    % (aecInVoice > 32/2^7 & anlpInVoice > anlpInNoise*8) rejects low-level
    % voice. By doing so, the detection can be more sensitive to lower
30    ratios
    % of near-end voice to total voice. Again, accepting conditions only
    when
    % the status gauge = 0 helps reject far-end single talk. The
    sensitivity
35    % to near-end speech is optimized by varying the threshold with
    % aecVoiceGainBase. The scale factor and offset is an empirical
    % compromise between avoiding false detection when the canceler is
    grossly
    % untrained and being as sensitive as possible to near-end speech after
40    % the canceler starts training.
    %
    % Method 3.
    % Comparing aecNearGain with aecVoiceGainBase detects near-end speech
    when
45    % echo suppression gets suddenly and consistently worse. This is
    because
    % aecVoiceGainBase does not react to sudden changes, and aecNearGain
    uses
    % a filter to ensure consistency in the detection. Because of the
50    filter
    % used for computing aecNearGain, conditions are more relaxed, which

```

```

% allows greater sensitivity to near-end speech. Such conditions
include
% having the status gauge < 0.3 rather than = 0, and having no
increased
5 % minimum absolute level for the voice. When far-end single talk
starts
% after the echo path has changed, aecNearGain can fall but still stay
% above aecVoiceGainBase for a short time. To avoid audible echo in
this
10 % case, near-end speech is only detected when aecNearGain is greater
than
% its last value.
if ((aecStatusGauge == 0) & (aecInEchoVoice < aecInVoice/4) & ...
(aecVoiceGainBase < 15155/32768)) | ...
15 ((aecStatusGauge == 0) & (aecInVoice > 32/2^7) & ...
(anlpInVoice > anlpInNoise*8) & ...
(aecNearRatio - aecVoiceGainBase*1.5 >= 9830/32768)) | ...
((aecNearGain - aecVoiceGainBase >= 6554/32768) & ...
(aecNearGain > aecNearGainLast)),
20 % Near-end speech has been detected.
% If this frame begins a new period of near-end speech (the hang time
% has expired and the last frame containing voice was only echo),
% then set the linear-echo-canceller echo gain equal to the baseline
% echo-gain estimate. This is done only at the beginning of a
25 near-end
% speech period so that the residual-echo suppresser has consistent
% attenuation during the near-end speech rather than increasing
% distortion as the baseline leaks upward while getting no updates.
if (anlpNearSpeechCount == 0) & (anlpNearSpeechFlag == 0),
30 aecEchoGain = aecVoiceGainBase;
end
% Set the flag indicating that the last frame with voice contained
% near-end speech. It will remain set during frames where voice is
not
35 % detected.
anlpNearSpeechFlag = 1;
% Restart the near-end-speech hang-time counter. When non-zero, it
will
% override anlpNearSpeechFlag to minimize distortion by the residual-
40 % echo suppresser of near-end speech during double talk or when the
% voice energy is too low to be detected.
% If this counter were to start only when far-end single talk was
% detected, there would be less cut-out of near-end speech during
double
45 % talk. However, the far-end person would hear echo every time
he/she
% started to speak after the near-end person spoke, even after a long
% pause.
% When there is a quick transition from near-end speech to far-end
50 % single talk (such as when the near-end person talks in the middle
of
% the far-end person's speech), the hang time will cause a short
period

```

```

% where the far-end person hears echo (at -25 dB). This artifact is
% worthwhile because of the significant reduction in cut-out during
% double-talk gained by the hang time.
anlpNearSpeechCount = 25;          % 25 * 20 ms = 500 ms
5 hangtime.
  end
  end

% If the echo-estimate voice energy is at least 15/16 the near-end voice
% energy, assume that this frame contains far-end echo speech only, and
10 % clear the near-end speech flag. If the near-end-speech hang-time
counter
% has already expired, the residual-echo suppresser will immediately go
to
% far-end single-talk mode. Otherwise, the residual-echo suppresser will
15 go
% to far-end single-talk mode when the near-end-speech hang-time counter
% expires. Note that near-end speech could be detected again while the
% counter is in progress, and then the residual-echo suppresser will stay
in
20 % near-end speech mode once the counter expires.
% This detection scheme compares 32-bit numbers and does not use any
% qualifiers based on the energy levels, the status gauge, or
measurements
% from previous frames (besides the noise estimates). Therefore, the
25 scheme
% is rather sensitive yet robust. Were this scheme to fail to detect
far-
% end speech, the far-end person would hear echo when he/she started to
% speak after the near-end person spoke, even after a long pause.
30 if (aecInEchoVoice > (aecInVoice - aecInVoice/16)),
    anlpNearSpeechFlag = 0;
end

% If the near-end-speech hang-time counter has expired, and the last
frame
35 % containing voice was only echo, then set the linear-echo-canceler echo
% gain equal to the last-measured linear-echo-canceler voice gain. This
% assumes that no near-end speech is present, so the residual-echo
% suppresser will attempt to suppress all of the voice. This test is
after
40 % that which clears anlpNearSpeechFlag so that aecEchoGain will reflect
the
% decision immediately.
if (anlpNearSpeechCount == 0) & (anlpNearSpeechFlag == 0),
    aecEchoGain = aecVoiceGain;
45 end

aecVoiceGainHist(frame) = aecVoiceGain;          % Save for debug
aecVoiceGainBase          % Echo for status
aecVoiceGainBaseHist(frame) = aecVoiceGainBase;  % Save for debug
aecEchoGain                % Echo for status
50 aecEchoGainHist(frame) = aecEchoGain;          % Save for debug

```

```

aecNearRatioHist(frame) = aecNearRatio;           % Save for debug
aecNearGainHist(frame) = aecNearGain;             % Save for debug
anlpNearSpeechCount      % Echo for status
anlpNearSpeechFlag       % Echo for status

5  % Calculate the loop echo gain up to the ANLP. The ANLP will attenuate
as
% needed to meet the total loop echo suppression goal for the system.
% ? is through the volume control.
% aecChanGain is from the loudspeaker to the microphone -- the channel.
10 % aecEchoGain is from the input to the output of the linear echo
canceler.
aecLoopEchoGain = aecChanGain * aecEchoGain; % Insert volume gain here.
% Quantize for Z.15 format
aecLoopEchoGain = floor(aecLoopEchoGain*32768)/32768

15 % Set the ANLP window size to capture the expected residual echo, but no
% more. This minimizes distortion on near-end voice and noise. The ANLP
% window size shrinks as the AEC improves its echo gain, so use
aecChanGain
% and aecEchoGain to control the window size. The volume-control gain is
20 % not used because the envelope-detector input for the ANLP comes after
the
% volume control. The echo gain estimates measure in an RMS sense, but
the
% ANLP needs to suppress the entire residual echo including peaks. Thus,
% use a peak-to-RMS factor multiplier (= 3). The dynamic range of
% anlpWindowGain is two because, when the linear echo canceler is grossly
% untrained, anlpWindowGain needs to be at least two to capture the echo
% within the window.
25 anlpWindowGain = min(2*ONE, 3 * aecChanGain * aecEchoGain);
% Quantize for Z1.14 format
30 anlpWindowGain = floor(anlpWindowGain*16384)/16384;

% If the VAD indicates that there only noise on the loudspeaker, then put
% the ANLP in a pass-through mode (gain = ONE). The VAD allows
high-quality
35 % near-end single talk because there is no distortion of the near-end
speech
% or noise. The VAD will often say that there is voice when there is
only
% noise, so other means are necessary to minimize distortion of the
40 near-end
% speech or noise in this case. Also, the VAD will infrequently say that
% there is only noise when there is a low level of voice. The problem is
% somewhat proportional to the noise level on the loudspeaker. However,
% having the linear echo canceler in the loop provides enough echo
45 % suppression to make the echo inaudible in these cases.
if speakerVad(frame) == 0, % Near-end single-talk condition.
    anlpEchoGain = ONE;
% Since speakerVad=1, the loudspeaker most likely has speech, but not
% necessarily. If near-end speech is detected, assume double talk. In
50 this

```



```

    % case, set the total loop echo suppression goal to -25 dB (1843/32768),
    and
    % set the ANLP gain to the needed echo suppression not provided in the
    rest
5    % of the loop. The ANLP gain is higher (the suppression is lower) and
    % distortion is reduced as the customer turns down the volume from full
    % scale. When the linear echo canceler is trained, the ANLP gain is
    % typically higher than -10 dB in this mode, so noise masking does not
    % improve the sound quality. The far-end user will hear some echo during
10    % double talk, but this artifact is preferable to cut-out or high
    distortion
    % of the near-end voice.
    elseif (anlpNearSpeechCount > 0) | (anlpNearSpeechFlag == 1), %
    Double-talk
15    anlpEchoGain = 1843/32768 / aecLoopEchoGain;
    % Since speakerVad=1 and near-end speech is not detected, assume there is
    % far-end single talk. The total loop echo suppression goal is -56 dB
    % (52/32768) so that echo is almost inaudible when both ends have quiet
    % backgrounds. As in double-talk mode, the ANLP gain is set to the
20    needed
    % echo suppression not provided in the rest of the loop, and the ANLP
    gain
    % is higher and distortion is reduced as the customer turns down the
    volume
25    % from full scale and as the linear echo canceler trains. However, the
    % gain needed to attenuate the echo to inaudibility below the noise floor
    % may be higher, so the higher of the two gains is used for the ANLP to
    % minimize distortion. Without the comfort noise, attenuating both the
    % echo and the noise by the same factor would not change the
30    signal-to-noise
    % ratio; so noise masking would not work. The comfort noise makes it
    such
    % that the noise at the input and output of the ANLP are the same level.
    % Therefore, the ANLP can attenuate the echo to the threshold of
35    audibility
    % below the noise floor, without distorting more than necessary. The
    square
    % root is taken of the ratio of noise to voice because these variables
    are
40    % in units of energy. Multiplying the desired echo-to-noise ration by
    the
    % actual noise-to-echo ratio will factor out the noise. What is left the
    % desired-to-actual echo ratio, which is the gain needed to mask the
    echo.
45    else
        % Far-end single-talk condition.
        ECHO_TO_NOISE_GOAL = 1/8; % -18 dB
        if (anlpInVoice > anlpInNoise),
            anlpEchoGain = max(52/32768 / aecLoopEchoGain, ...
                               ECHO_TO_NOISE_GOAL *
50    sqrt(anlpInNoise/anlpInVoice));
        else
            % Under noisy conditions, the gain doesn't exceed ECHO_TO_NOISE_GOAL,

```

```

    % even though theoretically it could for low-energy voice, because
the
    % noise estimate is too large as the noise level falls quickly (car
    % slows down). This causes echo to be heard when the ANLP echo gain
5 is
    % too high.
    anlpEchoGain = max(52/32768 / aecLoopEchoGain, ECHO_TO_NOISE_GOAL);
    end
10 anlpEchoGain = min(ONE, anlpEchoGain);
    % Quantize for Z.15 format
    anlpEchoGain = floor(anlpEchoGain*32768)/32768
    anlpGainHist(frame) = anlpEchoGain;          % Save for debug

    % Decrement the near-speech hang counter, if need be, so that it works
15 % independently of the loudspeaker VAD.
    anlpNearSpeechCount = max(0, anlpNearSpeechCount - 1);

    %
    % ANLP sample section
    %

20 % If the ANLP echo gain is ONE, the ANLP is inactive -- skip to save
MIPS.
    if (anlpEchoGain == ONE),
        % Keep the envelope detector running.
        for k = 1:FRAME_SIZE,
25            anlpRef = aecRef(FRAME_SIZE-k+1);
            anlpRefEnvelope = max(abs(anlpRef), 255/256 * anlpRefEnvelope);
            anlpRefEnvelope = floor(anlpRefEnvelope*2^31)/2^31;
        end
        % Update the variable used by the AC-center attenuator to be the same
30 as
        % what would result from processing the whole frame.
        anlpOutLast = uplinkAudio(FRAME_SIZE);
    else
        for k = 1:FRAME_SIZE,
35            % ANLP far-end ref = AEC ref. Using aecEchoEst instead or in
addition
            % gives no better results because aecEchoEst is rather unrelated to
the
            % residual echo. The AEC ref works equally well since
40            % it precedes the earliest echo contained in anlpIn. The long time
            % constant in the peak detector is a key to this ANLP, and it makes
            % close delay matching of anlpRef to the residual echo unnecessary.
            % The offset into the AEC ref delay line can be changed to compensate
            % for fixed delays in the echo path due to upsampling, downsampling,
45            % buffers, and/or minimum channel delay.
            anlpRef = aecRef(FRAME_SIZE-k+1);
            % Envelope detect (peak detect) anlpRef signal.
            % The exponential decay of the peak detector models the decay of the
            % reverberation in the car. The time constant is set to handle the
50 most

```

```

% slowly-decaying reverberation condition expected.
% A pole less than 255/256 results in echo getting through.
% A pole greater than 255/256 results in excess distortion to
near-end.
5   anlpRefEnvelope = max(abs(anlpRef), 255/256 * anlpRefEnvelope);
% anlpRefEnvelope should be 32 bits for storage.
% anlpRefEnvelope rounded to 16 bits would not decay lower than
%  $512/2^{15}$ . anlpRefEnvelope truncated to 16 bits would decay 1 bit
per
10  % sample when below  $256/2^{15}$ , and this is too fast (resulting in echo
% let through).
% Quantize for Z.31 format
anlpRefEnvelope = floor(anlpRefEnvelope*231)/231;
15  ONE.
% ANLP's Delta value is gain controlled by the AEC and limited to
anlpDelta = min(ONE, ...
                anlpWindowGain * floor(anlpRefEnvelope*32768)/32768);
% Quantize for Z.15 format
anlpDelta = floor(anlpDelta*32768)/32768;
20  anlpDeltaHist((frame-1)*FRAME_SIZE+k) = anlpDelta;      % Save for
debug
%
% Execute AC-center attenuator.
%
25  % The ANLP input is connected to AEC output via the noise suppresser.
anlpIn = uplinkAudio(k);
% If the input is below the window,
if (anlpOutLast - anlpIn) >= anlpDelta,
% Use all of signal outside window and attenuate signal within
30  window.
    anlpOutLast = anlpIn + anlpDelta - ...
                  anlpEchoGain * anlpDelta;
% Else if the input is above the window,
elseif (anlpIn - anlpOutLast) >= anlpDelta,
35  % Use all of signal outside window and attenuate signal within
window.
    anlpOutLast = anlpIn - anlpDelta + ...
                  anlpEchoGain * anlpDelta;
% Else the input is inside the window.
40  else,
    % Attenuate the signal.
    anlpOutLast = anlpEchoGain * (anlpIn - anlpOutLast) + anlpOutLast;
end
% Quantize for S.15 format
45  anlpOutLast = floor(anlpOutLast*32768)/32768;      % Save for next
time.
%
% Add comfort noise such that the ANLP output noise has the same
level
50  % and a similar spectrum as the car noise input to the ANLP.
%
```

```

    % Use random samples from a frame of captured noise from the car.
    % This produces white noise at the same power as the car noise even
if
    % the captured audio from the car accidentally contains voice.
5    anlpSeed = rem(48271 * anlpSeed, 2147483647);
    anlpComfortNoiseIn = anlpComfortNoise(1 + ...
        rem(anlpSeed, COMFORT_NOISE_SIZE));
    % Filter the white noise using the ARMA model discussed above.
    % The following is equivalent, assuming anlpArGain = 1 - anlpArCoef.
10    % anlpComfortNoiseOut = anlpArCoef * anlpComfortNoiseOutOld + ...
    %   anlpArGain * (anlpComfortNoiseIn + 0.8125*anlpComfortNoiseInOld);
    ma = anlpComfortNoiseIn + 0.8125*anlpComfortNoiseInOld; % Moving
Average
    maDiff = anlpComfortNoiseOutOld - ma;
15    % Quantize for S1.15 format
    % Rounding is necessary to avoid a bias on the comfort noise.
    % Add 2^(-17) to force the 1's complement floating point to act the
same
    % as 2's complement when rounding a negative number with a fraction
20    % of
    % exactly 0.5.
    maDiff = round(maDiff * 32768 + 2^(-17))/32768;
    anlpComfortNoiseOut = ma + anlpArCoef * maDiff;
    % Quantize for S.15 format
25    % Rounding is necessary to avoid a bias on the comfort noise.
    anlpComfortNoiseOut = round(anlpComfortNoiseOut*32768 +
2^(-17))/32768;
    anlpComfortNoiseOut = max(-1, min(ONE, anlpComfortNoiseOut));
    % Delay variables for next loop.
30    anlpComfortNoiseInOld = anlpComfortNoiseIn;
    anlpComfortNoiseOutOld = anlpComfortNoiseOut;
    % Limit the comfort noise to the window size.
    anlpComfortNoiseOut = min(anlpDelta, ...
        max(-anlpDelta, anlpComfortNoiseOut));
35    % Scale the comfort noise so that the ANLP output noise equals the
ANLP
    % input noise in level.
    anlpComfortNoiseOut = anlpComfortNoiseOut * (ONE - anlpEchoGain);
    % Quantize for S.15 format
40    anlpComfortNoiseOut = floor(anlpComfortNoiseOut * 32768) / 32768;
    % Add comfort noise to ANLP output signal.
    uplinkAudio(k) = max(-1, min(ONE, anlpOutLast +
anlpComfortNoiseOut));
    end
45    end

    % Save to the file outputs collected over a frame in integer format.
    fwrite(fidOut, [uplinkAudio; aecOut]*32768, 'int16');

    %disp(['aecInEnergy = ' dec2hex(aecInEnergy * ENERGY_SCALE)])
    %disp(['aecErrorShift = ' dec2hex(aecErrorShift+(aecErrorShift<0)*65536)])
50    %disp(['aecUpdate(1) = '
    dec2hex(aecUpdate(1)*32768+(aecUpdate(1)<0)*65536)])

```

```

%disp(['aecUpdate(2) = '
dec2hex(aecUpdate(2)*32768+(aecUpdate(2)<0)*65536)])
%disp(['aecUpdateNear(1) = ' ...
%       dec2hex(aecUpdateNear(1)*32768+(aecUpdateNear(1)<0)*65536)])
5 %disp(['aecUpdateNear(2) = ' ...
%       dec2hex(aecUpdateNear(2)*32768+(aecUpdateNear(2)<0)*65536)])
%disp(['aecUpdatePeak2 = ' dec2hex(aecUpdatePeak2*2^31)])
%disp(['aecOutEnergy = ' dec2hex(aecOutEnergy * ENERGY_SCALE)])
%disp(['aecRefEnergy = ' dec2hex(aecRefEnergy*128)])
10 %disp(['anlpInEnergy = ' dec2hex(anlpInEnergy * ENERGY_SCALE)])
%disp(['aecUpdateBase = ' dec2hex(aecUpdateBase*32768)])
%disp(['aecUpdatePeak = ' dec2hex(aecUpdatePeak*32768)])
%disp(['aecPeakToBase = ' dec2hex(aecPeakToBase*16)])
%disp(['aecStatusGauge = ' dec2hex(aecStatusGauge*32768)])
15 %disp(['aecInNoise = ' dec2hex(aecInNoise*2^31)])
%disp(['anlpInNoise = ' dec2hex(anlpInNoise*2^31)])
%disp(['anlpComfortNoise(1) = ' ...
%       dec2hex(anlpComfortNoise(1)*32768+(anlpComfortNoise(1)<0)*65536)])
%disp(['anlpComfortNoise(2) = ' ...
20 %       dec2hex(anlpComfortNoise(2)*32768+(anlpComfortNoise(2)<0)*65536)])
%disp(['anlpArCoef = ' dec2hex(anlpArCoef*2^31)])
%disp(['anlpArGain = ' dec2hex(anlpArGain*32768)])
%disp(['aecChanGainTrial = ' dec2hex(aecChanGainTrial*7)])
%disp(['aecChanGain = ' dec2hex(aecChanGain*32768)])
25 %disp(['aecVoiceGainBase = ' dec2hex(aecVoiceGainBase*32768)])
%disp(['aecVoiceGain = ' dec2hex(aecVoiceGain*32768)])
%disp(['aecEchoGain = ' dec2hex(aecEchoGain*32768)])
%disp(['aecDenom = ' dec2hex(aecDenom*8)])
%disp(['aecNumer = ' dec2hex(aecNumer*2^18)])
30 %disp(['aecUpdateGain = ' dec2hex(aecUpdateGain*32768)])
%disp(['aecCoef(1) = ' dec2hex(aecCoef(1)*32768+(aecCoef(1)<0)*65536)])
%disp(['aecCoef(2) = ' dec2hex(aecCoef(2)*32768+(aecCoef(2)<0)*65536)])
%disp(['anlpWindowGain = ' dec2hex(anlpWindowGain*16384)])
%disp(['anlpEchoGain = ' dec2hex(anlpEchoGain*32768)])
35 %disp(['anlpEchoGain = ']); 20*log10(anlpEchoGain)
%disp(['anlpComfortNoiseInOld = ' dec2hex(anlpComfortNoiseInOld*32768)])
%disp(['anlpComfortNoiseOutOld = ' dec2hex(anlpComfortNoiseOutOld*32768)])
%disp(['uplinkAudio(1) = ' ...
%       dec2hex(uplinkAudio(1)*32768+(uplinkAudio(1)<0)*65536)])
40 %disp(['uplinkAudio(2) = ' ...
%       dec2hex(uplinkAudio(2)*32768+(uplinkAudio(2)<0)*65536)])
%pause
%plot([abs(aecUpdate)/aecUpdateBase; ...
%       zeros(AEC_REF_TAPS - AEC_REF_TAPS - AEC_NEAR_TAPS, 1); ...
45 %       abs(aecUpdateNear)/aecUpdateBase])
%axis([0 AEC_REF_TAPS 0 aecPeakToBase+0.1])
%xlabel('Update element'),ylabel('Normalized update magnitude'),pause
%plot(20*log10(abs(aecCoef .* profile)),axis([0 AEC_COEF_TAPS -100 1]);
%xlabel('Coefficient number'),ylabel('Magnitude in dB'),pause

50 end

fclose(fidOut);

```

```

clear AEC_MAX_GAIN_THRESH AEC_BASELINE_THRESH AEC_STATUS_GAUGE_SCALER ONE
clear AEC_COEF_TAPS AEC_NEAR_TAPS AEC_REF_TAPS
clear FRAME_SIZE FRAME_BITS ENERGY_SCALE COMFORT_NOISE_SIZE
clear confirmedNoVoiceFlag
5 clear LRmatrix downlinkAudio uplinkAudio aecRef aecUpdate aecUpdateNear
   aecOut
   clear frame m k i anlpSeed aecEchoEst aecShiftPending aecErrorShift ASM T
   clear aecRefEnergy aecInEnergy aecOutEnergy anlpInEnergy aecEchoEstEnergy
   clear anlpIn anlpRef anlpRefEnvelope anlpDelta anlpOutLast
10 clear aecUpdatePeak2 aecUpdatePeak aecUpdateBase aecPeakToBase
   clear aecStatusGauge aecDenom aecNumer aecUpdateGain aecLoopEchoGain
   clear aecInNoiseStateVars aecInEchoNoiseStateVars anlpInNoiseStateVars
   clear aecInVoice anlpInVoice aecInEchoVoice aecNearVoice
   clear aecNearRatio aecNearGainLast anlpEchoGain
15 clear aecChanGainTrial anlpComfortNoiseIn anlpComfortNoiseOut
   clear aecPeakGain anlpWindowGain anlpDenom anlpArGain error quotient ma
   clear fidIn fidOut anlpNearSpeechCount anlpNearSpeechFlag
   ECHO_TO_NOISE_GOAL

% Estimate the noise frame energy based on the frame energy of a signal.

20 % A norm-based noise estimate provides a wider dynamic range with 32-bit
   % variables, and thus the operation remains consistent even at very low
   % noise
   % levels (such as the far end of test_65.raw). However, the norm-based
   % algo
25 % requires square-root and 32-bit-square operations (more MIPS). This
   % energy-
   % based algo has been optimized for quantization of low noise levels, and
   % it
   % does an acceptable job -- especially considering that noise level
30 % estimation
   % is not critical for very low noise levels (voice is always much larger
   % than
   % the noise). It also turns out that adaptation speeds can be made the
   % same.
35 % for the norm-based and energy-based algos. All pumping operations here
   % use
   % shifts -- just shift one more bit for the norm-based algo.
   %
40 % The goal is to update the estimate quickly when there is no speech in the
   % signal and slowly otherwise. Constant adaptation is needed to track the
   % car
   % noise as it changes rapidly. The voice-activity detector (VAD) output is
   % a
   % good start for determining when to update quickly, but it sometimes
45 % indicates no voice when there really is a low level of voice which is
   % much
   % higher than the noise. noVoiceFlag is the VAD output qualified by the
   % status gauge of the linear echo canceler to improve the reliability.
   % However, the status gauge does not differentiate noise from near-end
50 speech,

```

```

% double-talk, or tones. Therefore, the algorithm must to tolerant of some
% speech during the fast update period. Where the status gauge works well
% is
5 % during far-end single talk, which is where it is most important for the
% noise estimate to be accurate to avoid audible echo. The trick used here
% to
% reduce false deviations (primarily occurring during near-end speech) is
% to
10 % pump the estimate up and down at fixed rates based on comparison results
% rather than linearly filter the energy signal. Using a pump-up time
% constant that is 1/4 of the pump-down time constant biases the estimate
% toward the noise floor in spite of some voice. Pumping up and down by
% fixed
15 % increments instead of fixed time constants would result in a time
constant
% that changes with noise level.
%
% It is difficult to tell the difference between a rising noise floor and
% speech. The pump-up time constants here determines the tracking rate.
20 % However, the pump-up is necessarily slow to minimize false growth during
% speech. State machines are implemented to allow a faster increase in the
% noise estimate under certain conditions. If the state machine sees the
% signal energy greater than 8 times the noise estimate for at least 900 ms
% during no voice, it is assumed that the noise floor has increased, and
25 % the
% noise floor is set equal to the test noise estimate from test period.
% When the input noise is dominated by noise from the far end, blanking due
% to the AMPS in-band control channel or due to poor RF conditions will
% cause the noise floor to temporarily drop. The state machine attempts to
30 % restore the noise estimate after blanking. First confirm that the energy
% drop is between 5 and 25 frames long. Then confirm that the energy
returns
% back to the original level. If the energy remains within a window around
% the original noise level, a relatively short confirmation period is
35 % needed.
% If the energy jumps much higher than the original level, then voice could
be
% occurring, and a longer confirmation period is needed to ensure that the
% noise floor has not dropped.
40 %
% Definition of noiseStateVars array:
% (1) = noise estimate
% (2) = noise estimate from before blanking or test noise estimate
% (3) = state variable/counter
45 %
function [confirmedNoVoiceFlag, noiseStateVars] = ...
    estnoise(inEnergy, noVoiceFlag, noiseStateVars)

% Define the needed constants in the same way as the parent routine.

% The number of samples in the update integration period.
50 FRAME_SIZE = 160;

```

```

% Number of bits to right shift values accumulated over a frame of samples.
FRAME_BITS = ceil(log2(FRAME_SIZE));

% Scale factor to quantize energies to 32 bits (Z8.23 format
w/FRAME_SIZE=160)
5 ENERGY_SCALE = 2^(31-FRAME_BITS);

% If the parent routine has initialized the noise estimate,
if noiseStateVars(1) == FRAME_SIZE,
%
% Take actions to speed up adaptation for the beginning of a call.
%
10 % Set the flag so that the comfort noise will initialize with this level.
confirmedNoVoiceFlag = 1;
% Start with the noise estimate equal to the energy.
noiseStateVars(1) = inEnergy;
15 else
%
% Update the noise estimate.
%
% By default, clear the flag so as to indicate no comfort noise training.
20 confirmedNoVoiceFlag = 0;
%
% The following bias test was performed in Matlab on white noise:
% >>n=160*3000;noise=zeros(1,m);
% >>seed=1;for i=1:n,[noise(i) seed]=noisegen(seed);end
25 % >>m=3000;for i=1:m,noiseEnergy(i)=sum(noise(160*(i-1)+1:160*i).^2);end
% >>ne=100;for i=1:m,if ne>noiseEnergy(i),
% >>ne=ne-ne/16;else,ne=ne+ne/64;end,neh(i)=ne;end
% >>plot(neh) % neh is the noise estimate history.
% >>axis([0 20 0 100]) % Notice that neh settles after 10 frames.
30 % >>1/(sum(neh(11:m))/(m-10)/(sum(noiseEnergy)/m)-1)
% ans = -14.2082
% This indicates that the noise estimate has a bias factor of about
-1/14.
% To simplify arithmetic, subtract a factor of 1/16 from the noise
35 estimate
% for comparison purposes to restore the bias.
% In assembly, right shift before subtraction for bit exactness.
noiseBiased = noiseStateVars(1) - noiseStateVars(1)/16;
% Quantize to 32 bits
40 noiseBiased = ceil(noiseBiased*ENERGY_SCALE)/ENERGY_SCALE;
%
% If the VAD and status gauge indicate no voice,
if noVoiceFlag,
%
45 % Pump the noise estimate quickly.
%
% If the noise estimate is too high,
if noiseBiased > inEnergy,
%
50 % Pump the noise estimate down.
%

```



```

    % To speed up tracking with a decreasing noise floor, it was tried to
    % set the noise estimate equal to the signal energy immediately if
the
    % noise estimate was too high by a factor of 8. However, this made
5 the
    % bias very negative when activated, and it created problems with
false
    % detections. The time constant is so short that pumping the noise
    % estimate down instead works quite well.
10
    % To minimize the deviation, don't pump down when the noise estimate
    % equals the signal energy.
    %
    % Quantization causes a minimum change of 1/ENERGY_SCALE, except at
15 0.
    % In assembly, negate before right shift for bit exactness.
    noiseStateVars(1) = noiseStateVars(1) - noiseStateVars(1)/16;
    % Set the flag for comfort noise training. Requiring that the input
    % energy be lower than the noise estimate improves the probability
20 that
    % comfort noise is not updated during voice.
    confirmedNoVoiceFlag = 1;
    else
    %
25
    % Pump the noise estimate up.
    %
    % Use a minimum increment to avoid getting stuck near zero.
    % Adding min(noiseStateVars(1)/64, 1/ENERGY_SCALE) is not used
instead
30
    % because it results in slower adaptation to suddenly increased
noise.
    noiseStateVars(1) = noiseStateVars(1) + noiseStateVars(1)/64 + ...
        1/ENERGY_SCALE;
    end
35
    else
    %
    % Pump the noise estimate slowly.
    %
    % Adaptation is not stopped during voice because of the importance of
40
    % accurately tracking a decreasing noise floor. Over-estimation of the
    % noise causes under-estimation of the voice energy. This has more of
an
    % impact on the NLP input than the linear echo canceler input due to
the
45
    % lower far-end voice energy. Thus, the result is insufficient echo
    % suppression. So it is better to under-estimate than over-estimate
the
    % noise.
    %
50
    % If the noise estimate is too high,
    if noiseBiased > inEnergy,
    %
    % Pump the noise estimate down.

```

```

    %
    % Quantization causes a minimum change of 1/ENERGY_SCALE, except at
0.    %
    % In assembly, negate before right shift for bit exactness.
5    noiseStateVars(1) = noiseStateVars(1) - noiseStateVars(1)/64;
    else
    %
    % Pump the noise estimate up.
    %
10    % At first glance, it may seem that only pumping down during voice is
    % necessary to accurately track a decreasing noise floor. However,
this    % will cause the bias to become too strongly negative. The pump up
rate    % was empirically determined to be the fastest possible while not
15    % allowing too much of a false change during voice. This turns out
to    % be very slow since voice can last for several seconds between
pauses.
20    % Don't use a minimum increment or else the ramp up will be too
    % large for low noise levels during voice.
    noiseStateVars(1) = noiseStateVars(1) + noiseStateVars(1)/1024;
    end
    end
25    % Quantize to 32 bits
    noiseStateVars(1) = floor(noiseStateVars(1)*ENERGY_SCALE)/ENERGY_SCALE;
    %
    % State machine for AMPS blanking and noise jump tracking.
    % Don't execute at initialization.
30    %
    % If the state machine is in the idle state,
    if noiseStateVars(3) == 0,
    % If the signal energy has significantly dropped below the noise
estimate,
35    % and if quantization of inEnergy does not give false results,
    if noiseStateVars(1) > max(8*inEnergy, 8/ENERGY_SCALE),
    % Store the noise estimate for the state machine.
    noiseStateVars(2) = noiseStateVars(1);
    % Start the state machine to look for blanking.
40    noiseStateVars(3) = 1;
    % If the VAD and status gauge indicate no voice, and the signal energy
is    % significantly higher than the noise estimate, and if quantization of
    % the noise estimate does not give false results,
45    elseif noVoiceFlag & ...
        (inEnergy > max(8*noiseStateVars(1), 8/ENERGY_SCALE)),
    % Initialize the test noise estimate.
    noiseStateVars(2) = inEnergy;
    % Start the state machine to look for a noise jump.
50    noiseStateVars(3) = -1;
    end
    % Else if the state machine is looking for a noise jump,
    elseif noiseStateVars(3) < 0,

```

```

% If the VAD and status gauge continue to indicate no voice, and if the
% signal energy remains significantly higher than the noise estimate,
if noVoiceFlag & (inEnergy > 8*noiseStateVars(1)),
    % Bias the test noise estimate just like the regular one.
5    % In assembly, right shift before subtraction for bit exactness.
    noiseBiased = noiseStateVars(2) - noiseStateVars(2)/16;
    % Quantize to 32 bits
    noiseBiased = ceil(noiseBiased*ENERGY_SCALE)/ENERGY_SCALE;
    % If the test noise estimate is too high,
10    if noiseBiased > inEnergy,
        % Pump the test noise estimate down.
        % Quantization causes a minimum change of 1/ENERGY_SCALE.
        % In assembly, negate before right shift for bit exactness.
        noiseStateVars(2) = noiseStateVars(2) - noiseStateVars(2)/16;
15    else
        % Pump the test noise estimate up.
        noiseStateVars(2) = noiseStateVars(2) + noiseStateVars(2)/64;
    end
    % Quantize to 32 bits
20    noiseStateVars(2) =
    floor(noiseStateVars(2)*ENERGY_SCALE)/ENERGY_SCALE;
    % Decrement the state variable which also acts as a counter.
    noiseStateVars(3) = noiseStateVars(3) - 1;
    % If the signal energy has remained significantly higher than the
25    noise
    % estimate for a sufficient period,
    % (45 frames are needed for the echo of test track s_top10_1.raw)
    if noiseStateVars(3) == -45,
        % Jump the noise estimate to the test noise estimate.
30        noiseStateVars(1) = noiseStateVars(2);
        % Reset the state machine back to the idle state.
        noiseStateVars(3) = 0;
    end
    else,
35        % Reset the state machine back to the idle state.
        noiseStateVars(3) = 0;
    end
    % Else the state machine is looking for blanking.
    else,
40        % Increment the state variable.
        noiseStateVars(3) = noiseStateVars(3) + 1;
        % States 1-100 count the number of frames in the alleged blanking
        period.
        if noiseStateVars(3) < 101,
45            % If the energy goes back high,
            if 8*inEnergy > noiseStateVars(2),
                % If the blanking is less than 5 frames,
                if noiseStateVars(3)-1 < 5,
                    % Either the detection was false or it is not worth restoring the
50                    % noise estimate. Put the state machine back into idle state.
                    noiseStateVars(3) = 0;
                else
                    % Voice has occurred or the noise has returned after blanking.

```

```

        % Set the state variable to 101 to start the next phase.
        noiseStateVars(3) = 101;
    end
    % If the count of low energy frames is too long,
5    elseif noiseStateVars(3)-1 == 25,
        % Blanking did not occur -- the noise floor dropped instead.
        % Put the state machine back into idle state.
        noiseStateVars(3) = 0;
    end
10    % The state machine has detected the end of the alleged blanking
    period.
    % Regardless of whether the state machine is counting frames of voice
    or
    % noise, first check if the energy goes back low.
15    elseif 8*inEnergy < noiseStateVars(2),
        % Blanking did not occur -- the noise floor dropped instead.
        % Put the state machine back into idle state.
        noiseStateVars(3) = 0;
    % States 101-200 count the number of frames of voice or noise following
    % the alleged blanking period.
20    elseif noiseStateVars(3) < 201,
        % If the energy is goes very high,
        if inEnergy > noiseStateVars(2)*8,
            % Assume that this is voice.
25            % Set the state variable to 201 to start the next phase.
            noiseStateVars(3) = 201;
            % If there is a sufficient count of frames where the maximum and
            minimum
            % energy is close to the saved noise estimate,
30            elseif noiseStateVars(3)-101 == 10,
                % The blanking is confirmed.
                % Restore the noise estimate to that before the blanking.
                noiseStateVars(1) = noiseStateVars(2);
                % Put the state machine back into idle state.
35                noiseStateVars(3) = 0;
            end
            % States 201-300 count the number of frames of voice following
            % the alleged blanking period.
            %
40            % If there is a sufficient count of frames where the minimum
            % energy is close to the saved noise estimate,
            elseif noiseStateVars(3)-201 == 50,
                % The blanking is confirmed.
                % Restore the noise estimate to that before the blanking.
45                noiseStateVars(1) = noiseStateVars(2);
                % Put the state machine back into idle state.
                noiseStateVars(3) = 0;
            end
        end
50    end
end

```

Those skilled in the art will appreciate that the present invention is not limited to the specific exemplary embodiments which have been described herein for purposes of illustration. For example, the various operational blocks of the disclosed embodiments are conceptual in nature. Actual implementation of the functions of such blocks can be accomplished using a variety of techniques. Furthermore, each exemplary system can be implemented, for example, using multiple standard digital signal processing chips, a single application-specific integrated circuit, or an appropriately configured computer.

Note also that, although the exemplary embodiments have been described in the context of acoustic echo canceling, the teachings of the present invention are equally applicable in the context of network echo canceling (e.g., where the near-end user is a landline user and the far-end user is a mobile user). Further, certain aspects of the present invention are applicable to communications systems generally and are not limited to echo suppression systems. Thus, the scope of the invention is defined by the claims which are appended hereto, rather than the foregoing description, and all equivalents which are consistent with the meaning of the claims are intended to be embraced therein.

CLAIMS:

1. An echo suppression device, comprising:
an echo suppressor configured to attenuate a first communications signal
in order to suppress an echo component thereof, said echo component resulting from a
5 reflection of a second communications signal; and
a voice activity detector coupled to said echo suppressor and configured
to provide an indication of whether said second communications signal includes a
speech component,
wherein said echo suppressor is deactivated when said voice activity
10 detector indicates that said second communications signal does not include a speech
component.
2. An echo suppression device according to claim 1, wherein said echo
suppressor is deactivated via a bypass.
3. An echo suppression device according to claim 1, wherein said echo
15 suppressor is deactivated by adjusting the echo suppressor to provide a level of
attenuation which is sufficient to prevent howling.
4. An echo suppression device according to claim 1, wherein said echo
suppressor is a single-throw switch and wherein said echo suppressor is deactivated by
closing said single-throw switch.
- 20 5. An echo suppression device according to claim 1, wherein said echo
suppressor is a clipping suppressor and wherein said echo suppressor is deactivated by
setting a clipping threshold of said echo suppressor to zero.

6. An echo suppression device according to claim 1, wherein said echo suppressor is a scaling suppressor and wherein said echo suppressor is deactivated by setting a scale factor of said echo suppressor to unity.

5 7. An echo suppression device according to claim 1, wherein said echo suppressor is a hybrid suppressor and wherein said echo suppressor is deactivated by setting a clipping threshold of said echo suppressor to zero and by setting a scale factor of said echo suppressor to unity.

8. An echo suppression device, comprising:
an echo suppressor configured to attenuate a first communications signal
10 in order to suppress an echo component thereof, wherein said echo component results from a reflection of a second communications signal, and wherein said echo suppressor multiplies at least a portion of said first communications signal by an attenuation factor;
and
a gain control processor configured to detect a prevailing condition of
15 said communications signals and to adjust said attenuation factor based on said detected prevailing condition when said echo suppressor is active.

9. An echo suppression device according to claim 8, wherein said gain control processor includes a voice activity detector configured to detect whether said first communications signal includes a near-end speech component, and wherein said
20 gain control processor adjusts said attenuation factor based upon whether said voice activity detector detects said near-end speech component.

10. An echo suppression device, comprising:
an echo suppressor configured to attenuate a first communications signal
in order to suppress an echo component thereof, said echo component resulting from a
25 reflection of a second communications signal; and

a gain control processor configured to detect a ratio of voice to noise in said first communications signal, wherein said gain control processor adjusts a level of attenuation provided by said echo suppressor based upon said ratio.

5 11. An echo suppression device according to claim 10, wherein said ratio of voice to noise is a ratio of echo to noise.

 12. An echo suppression device according to claim 10, wherein said gain control processor deactivates said echo suppressor when said ratio is at or below a predefined threshold.

10 13. An echo suppression device according to claim 12, wherein said echo suppressor is deactivated via a bypass.

 14. An echo suppression device according to claim 12, wherein said gain control processor deactivates said echo suppressor by adjusting the echo suppressor to provide a level of attenuation which is sufficient to prevent howling.

15 15. An echo suppression device according to claim 12, wherein said echo suppressor is a single-throw switch and wherein said echo suppressor is deactivated by closing said single-throw switch.

 16. An echo suppression device according to claim 12, wherein said echo suppressor is a clipping suppressor and wherein said echo suppressor is deactivated by setting a clipping threshold of said echo suppressor to zero.

20 17. An echo suppression device according to claim 12, wherein said echo suppressor is a scaling suppressor and wherein said echo suppressor is deactivated by setting a scale factor of said echo suppressor to unity.

FIG. 1A

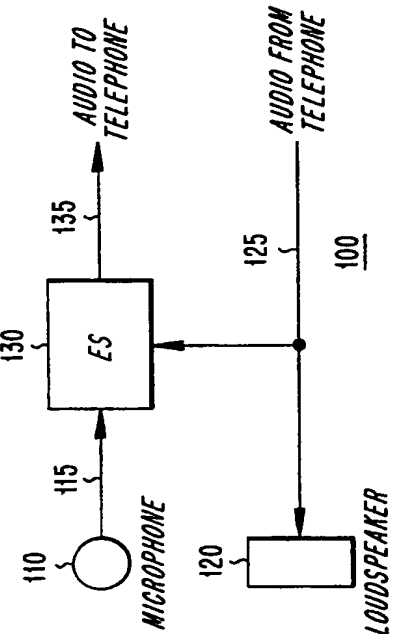
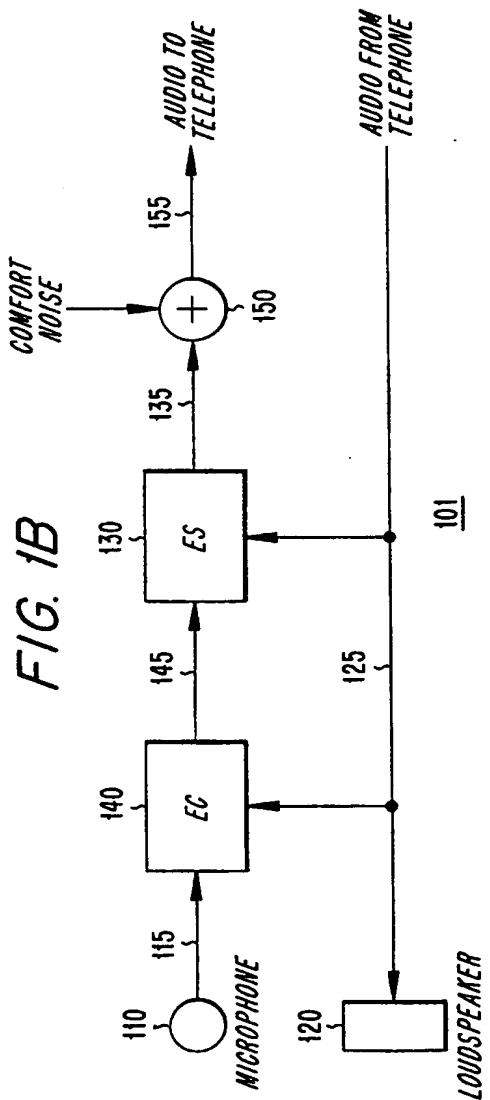


FIG. 1B



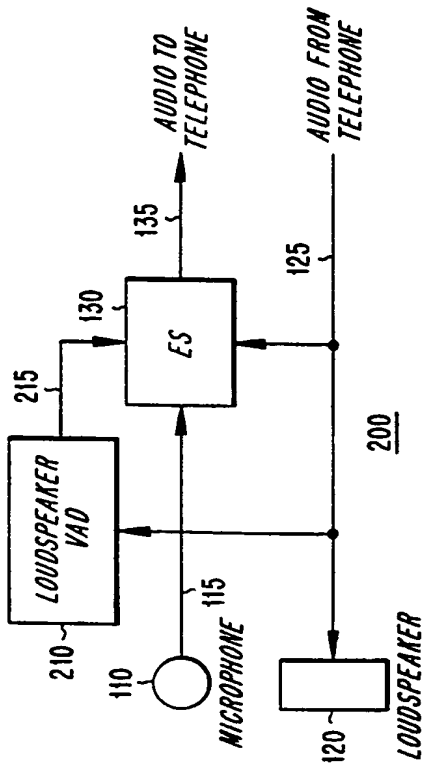


FIG. 2

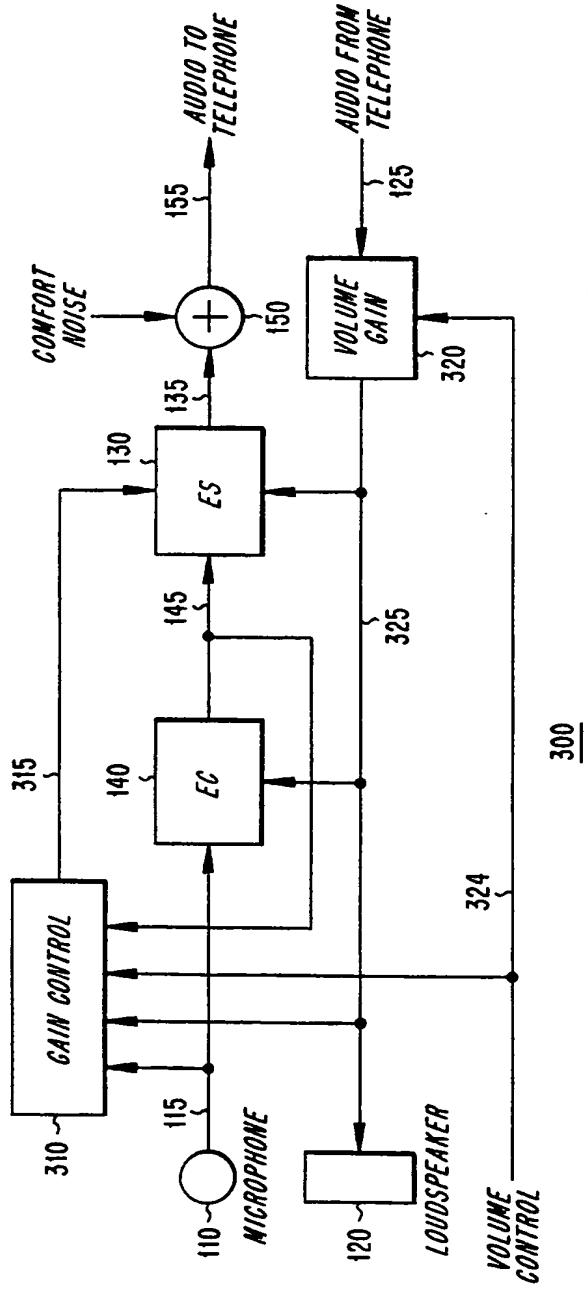


FIG. 3

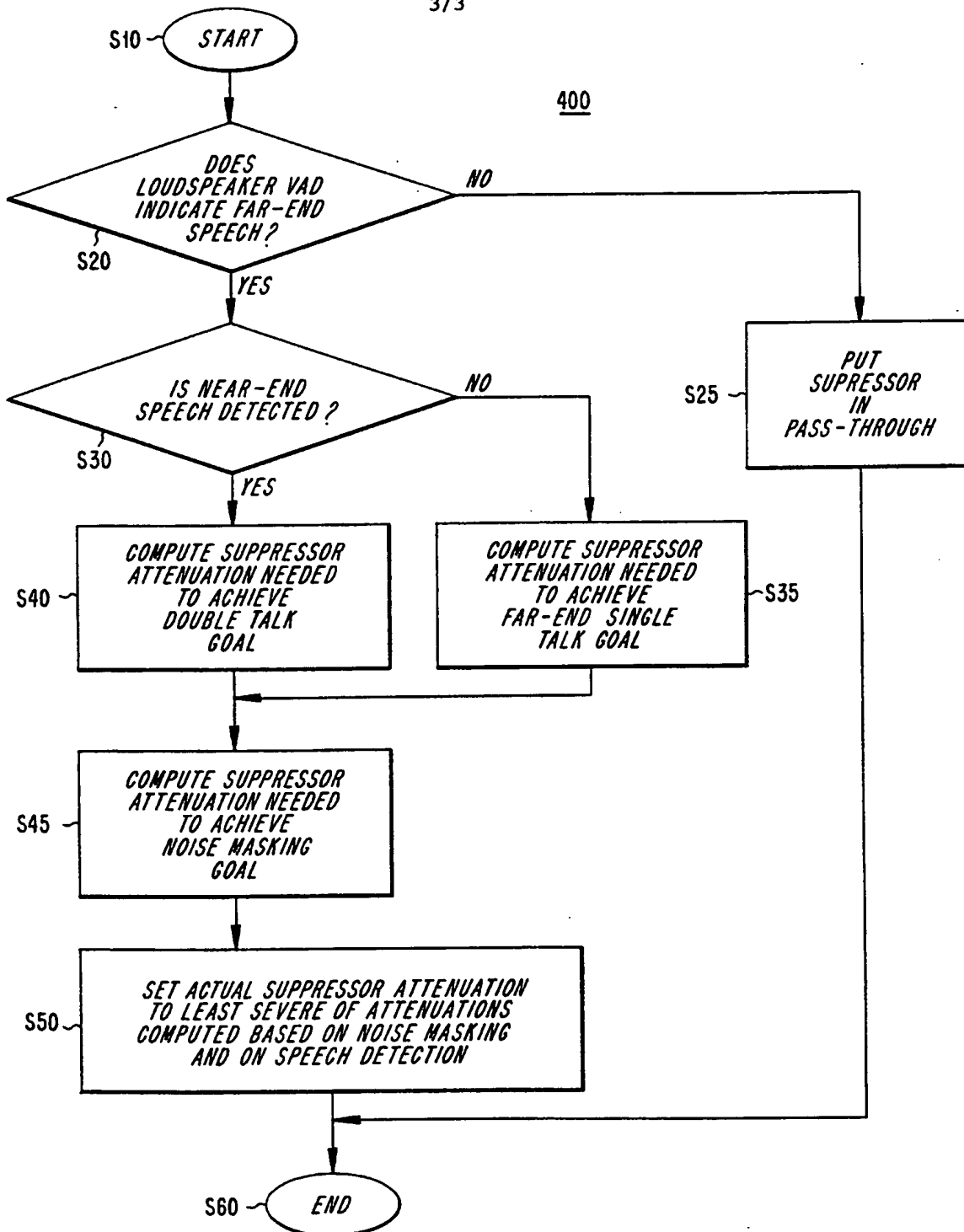


FIG. 4

INTERNATIONAL SEARCH REPORT

In tional Application No
PCT/US 98/26231

A. CLASSIFICATION OF SUBJECT MATTER

IPC 6 H04M9/08

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 H04M H04B

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X A	US 5 619 566 A (FOGEL ELIEZER) 8 April 1997 see column 4, line 4 - line 23 see column 6, line 11 - line 16 see figures 2-4 ---	1,3 8-10,12, 14
X	US 5 022 074 A (NICHOLAS DAVID C) 4 June 1991 see column 3, line 28 - line 51 see column 4, line 52 - line 66 see column 6, line 60 - column 7, line 13 see figure 1 --- -/--	1,3

☒ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

* Special categories of cited documents :

"A" document defining the general state of the art which is not considered to be of particular relevance

"E" earlier document but published on or after the international filing date

"L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)

"O" document referring to an oral disclosure, use, exhibition or other means

"P" document published prior to the international filing date but later than the priority date claimed

"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

"X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

"Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

"&" document member of the same patent family

Date of the actual completion of the international search

22 March 1999

Date of mailing of the international search report

31/03/1999

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
Fax: (+31-70) 340-3016

Authorized officer

Vaucois, X

INTERNATIONAL SEARCH REPORT

International Application No
PCT/US 98/26231

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT		
Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	US 5 475 731 A (RASMUSSEN JIM A J) 12 December 1995 cited in the application see column 2, line 35 - line 51 see column 7, line 59 - column 8, line 15 see figure 7 -----	1,5,16
A	GB 2 295 300 A (MOTOROLA ISRAEL LTD) 22 May 1996 see page 8, line 10 - line 14 -----	1
A	EP 0 739 123 A (NIPPON ELECTRIC CO) 23 October 1996 see column 3, line 58 - column 4, line 6 see figure 1 -----	1
P,X	WO 98 30009 A (ERICSSON GE MOBILE INC) 9 July 1998 cited in the application see page 5, line 7 - line 18 see page 9, line 17 - line 29 see page 11, line 30 - page 12, line 16 see page 13, line 14 - line 31 see figures 1,6 -----	1,5

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No
PCT/US 98/26231

Patent document cited in search report		Publication date	Patent family member(s)		Publication date
US 5619566	A	08-04-1997	GB 2281680 A		08-03-1995
			CN 1114122 A		27-12-1995
			EP 0671099 A		13-09-1995
			WO 9506382 A		02-03-1995
US 5022074	A	04-06-1991	NONE		
US 5475731	A	12-12-1995	AU 1560895 A		01-08-1995
			WO 9519085 A		13-07-1995
GB 2295300	A	22-05-1996	NONE		
EP 0739123	A	23-10-1996	JP 8293820 A		05-11-1996
			AU 5079296 A		31-10-1996
WO 9830009	A	09-07-1998	AU 7893798 A		31-07-1998